

TSTE25 Power Electronics

Lecture 2

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ISY/FS

Outline

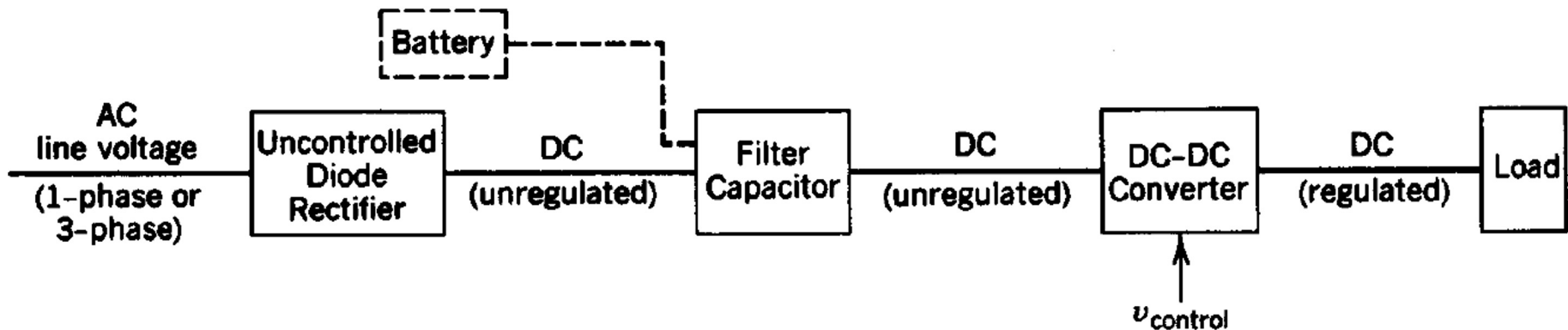
- DC-DC Converter
 - Step-down (Buck)
 - [Step-up \(Boost\)](#)
 - [Buck/Boost \(Self-study for interested\)](#)

Lecture 2

DC/DC (Buck) step-down converter

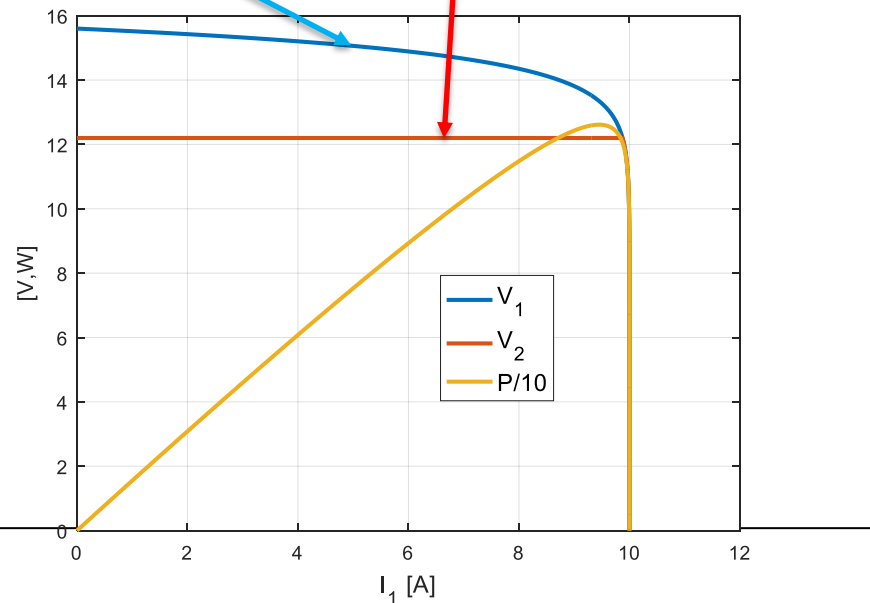
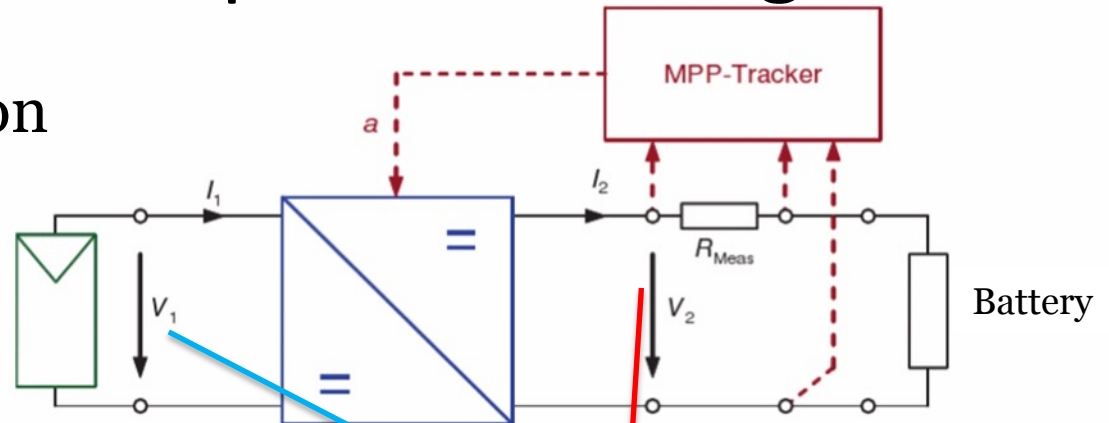
Basic use of DC-DC converter

- Unregulated DC input, controlled DC output
 - Regulated DC may be larger or smaller than the unregulated DC voltage
 - Input to the DC-DC converter may vary a lot



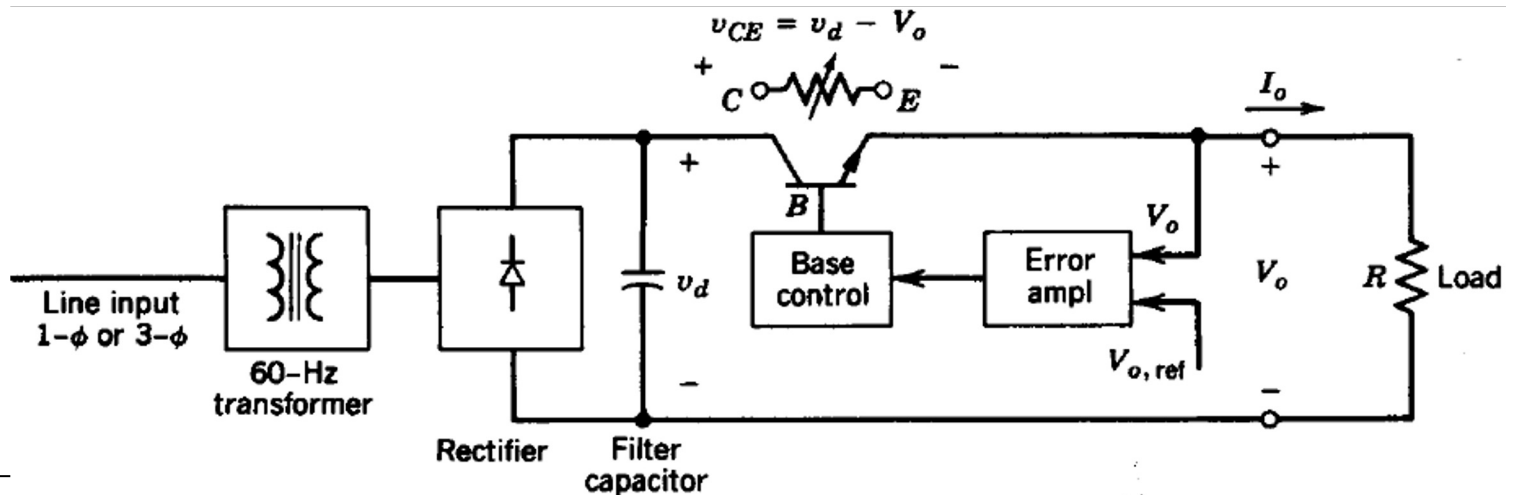
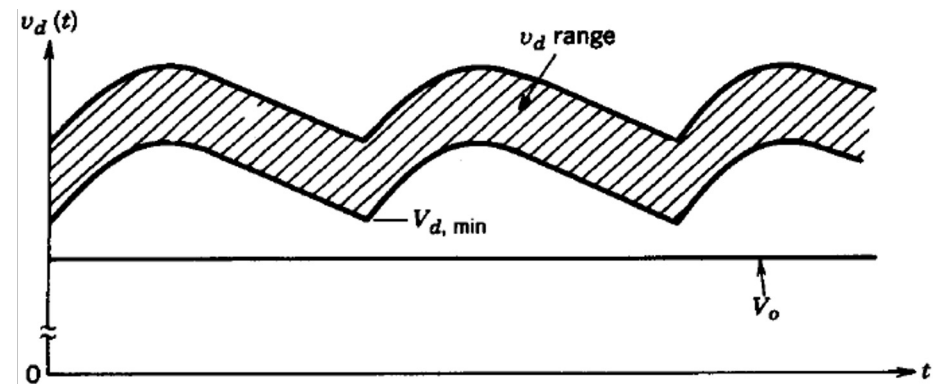
PV-panel maximum power tracking

- DC/DC conversion to optimize PV-panel voltage for maximum power
- Constant (12V) across battery load
- Variable PV-panel voltage (12-16 V)



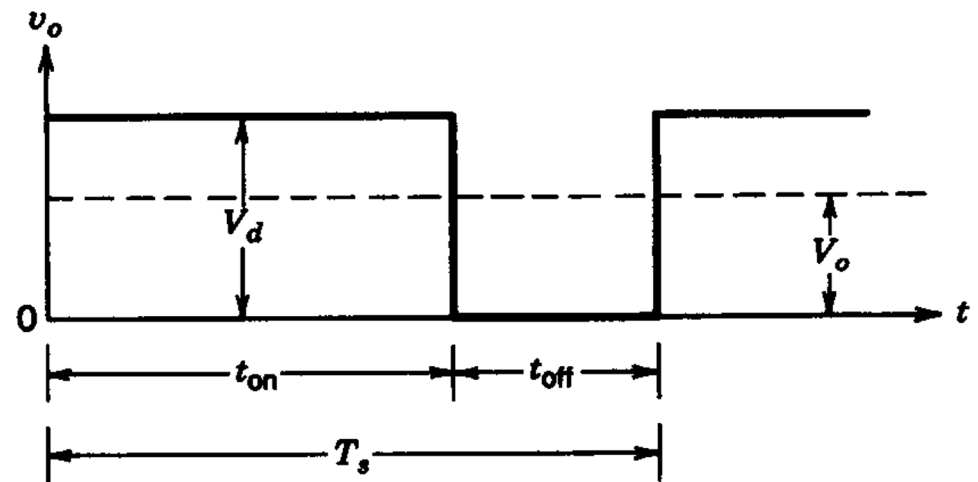
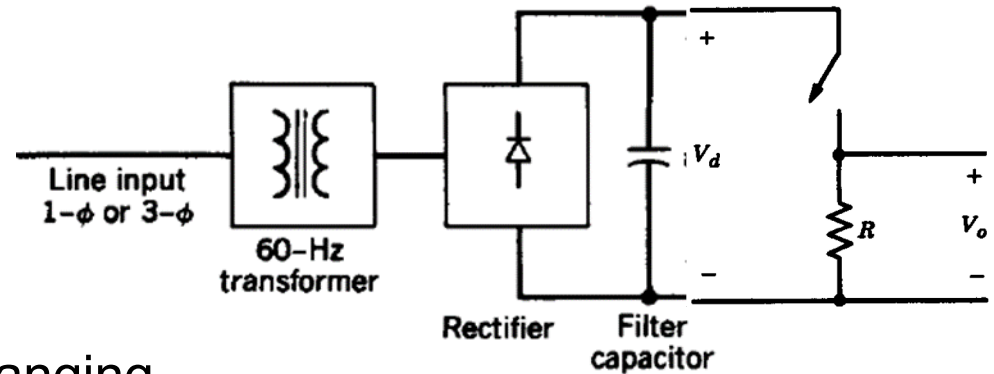
Linear power supply

- Bulky transformer
 - Low frequency
- Poor efficiency
 - 30 – 60 %
- Low EMI



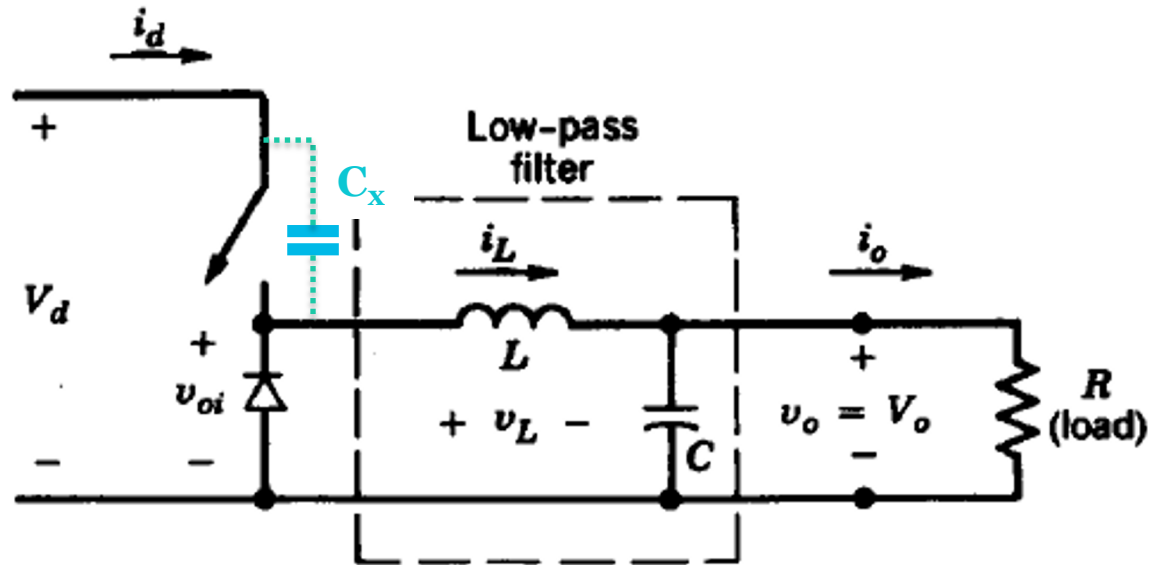
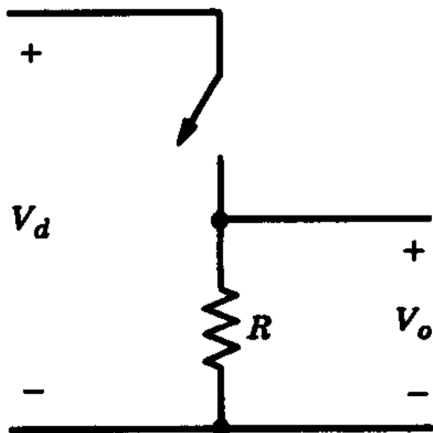
Step down converter principle

- $V_d > V_o$
- T_s constant, t_{on} and t_{off} changing
- Large ripple in V_o



Step-down (Buck) converter

- Add filter to reduce ripple voltage
- Diode added to protect switch
 - Without diode: V_L very high when switch opens with current in L due to very high di/dt .
 - Parasitic capacitances C_x would be charged by the inductor current



DC/DC-converter control

- Pulse width modulation, PWM, to control switching
- Switching frequency, f_s , the repetition rate of turn-on/off

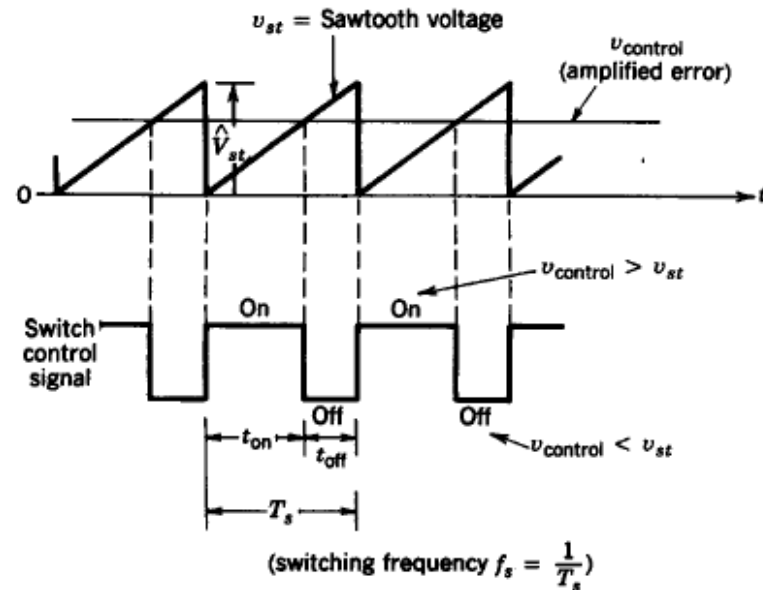
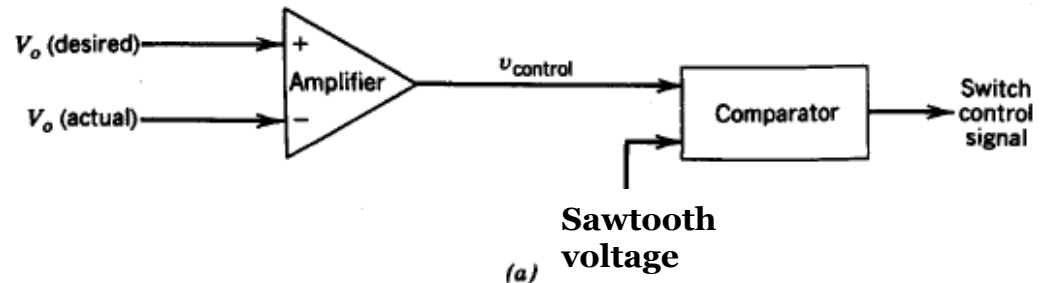


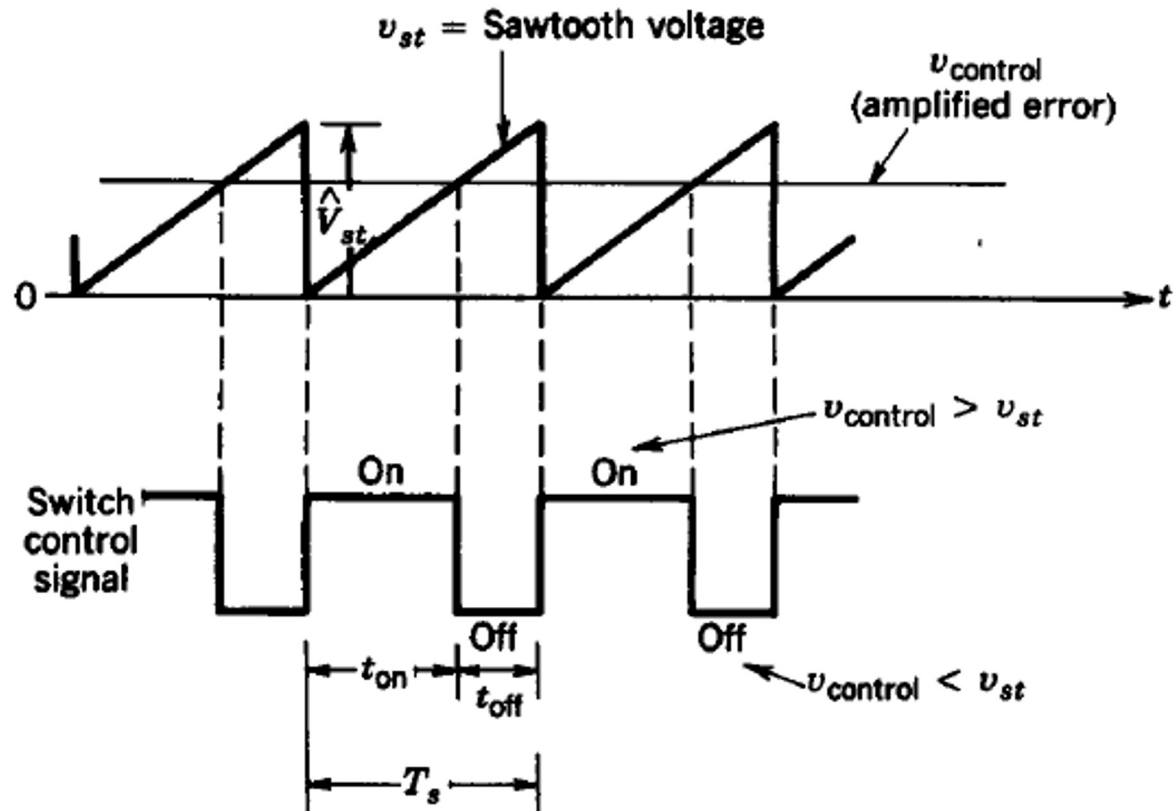
Figure 7-3 Pulse-width modulator: (a) block diagram; (b) comparator signals.

PWM waveform, duty cycle

- Switch duty cycle (duty ratio)

$$D = \frac{t_{on}}{T_s} = \frac{v_{control}}{\hat{V}_{st}}$$

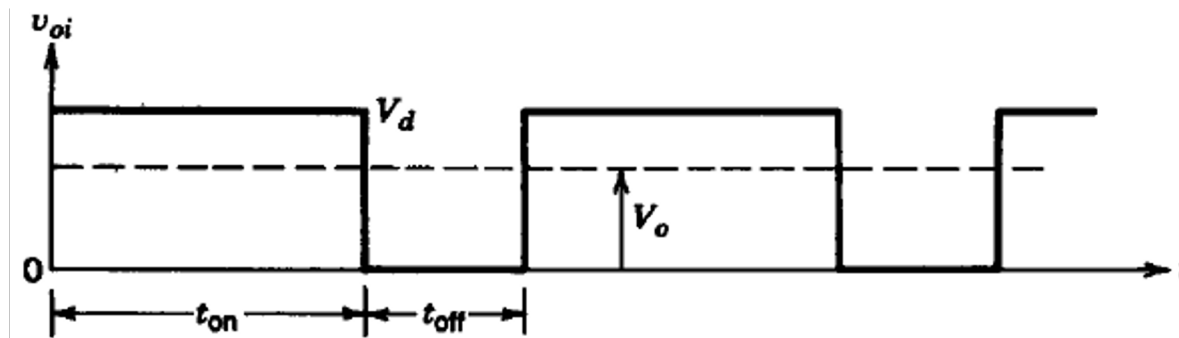
$$0 < D < 1$$



Step-down converter waveforms

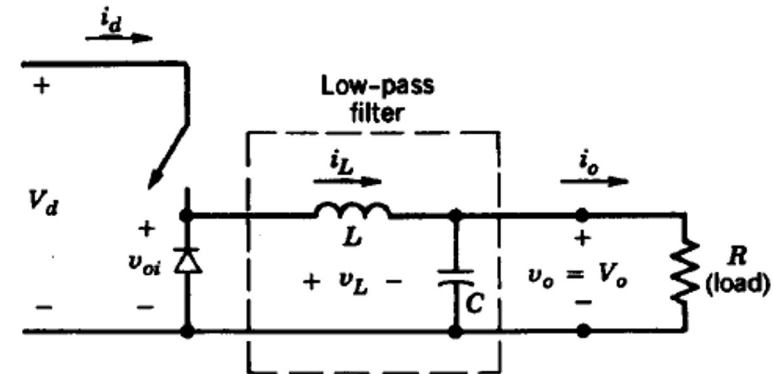
- $T_s = t_{on} + t_{off}$
- Average output voltage

$$V_o = \frac{t_{on}}{T_s} V_d = D \cdot V_d$$



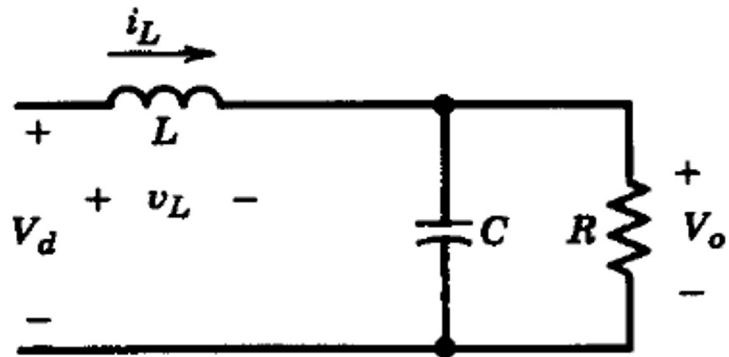
Current Conduction modes

- Average i_L equals i_o
- Two current conduction modes (i_L)
 - Continuous current conduction
 - Non-continuous current conduction
- Converter characteristics different depending on mode
- Both modes can be supported by a converter
 - Mode applicable is depending on load

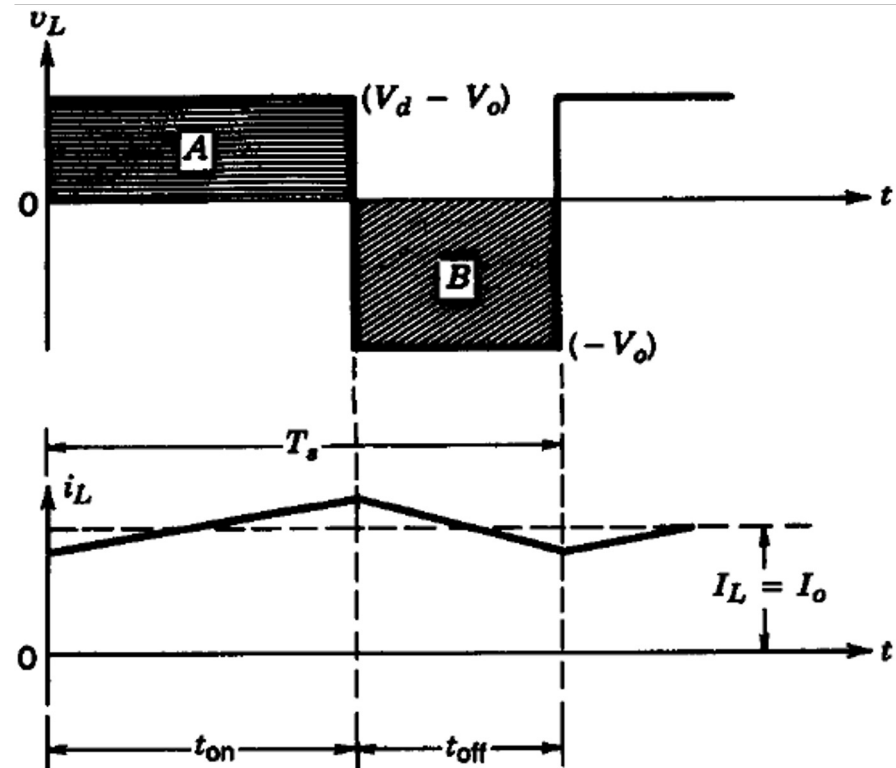
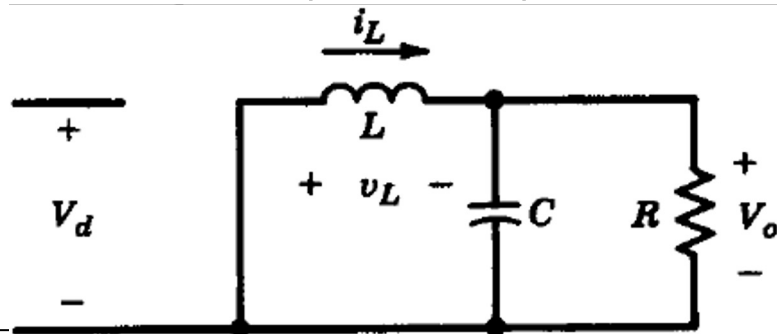


Continuous Conduction mode

- Switch on (diode off)

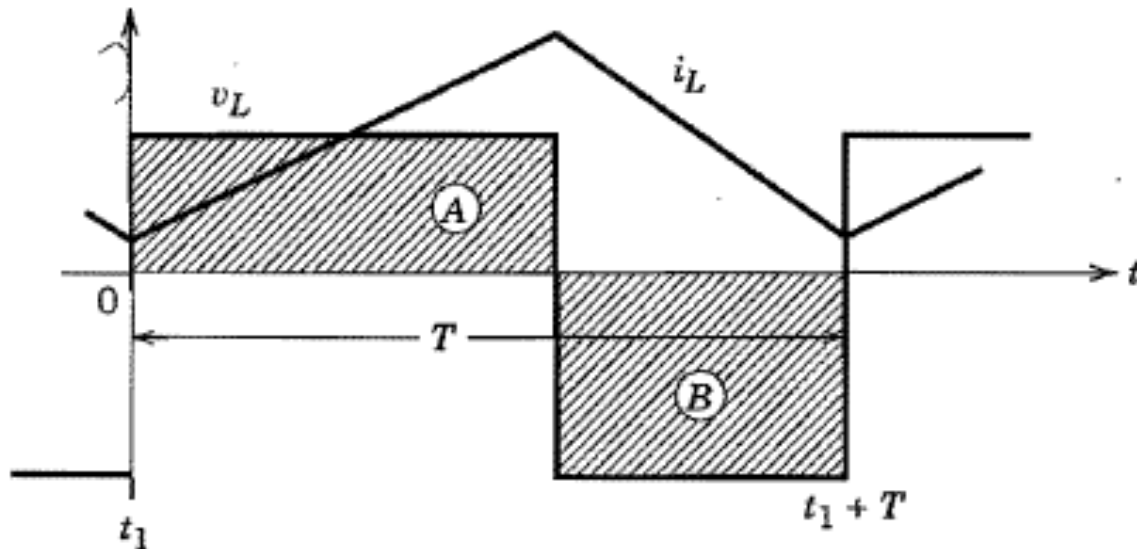
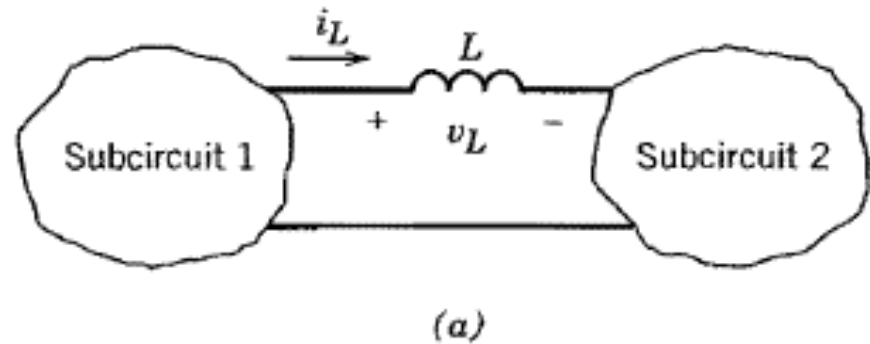


- Switch off (diode on)



Inductor a current storage

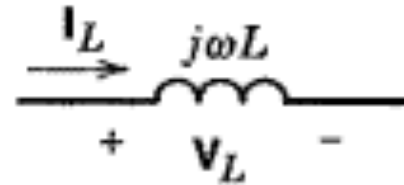
- Voltage is required to change the current
- $v_L > 0$: i_L increase
- $v_L < 0$: i_L decrease
- Steady state:
 - Constant average current
 - Voltage-time area $A=B$
 - Average voltage=0



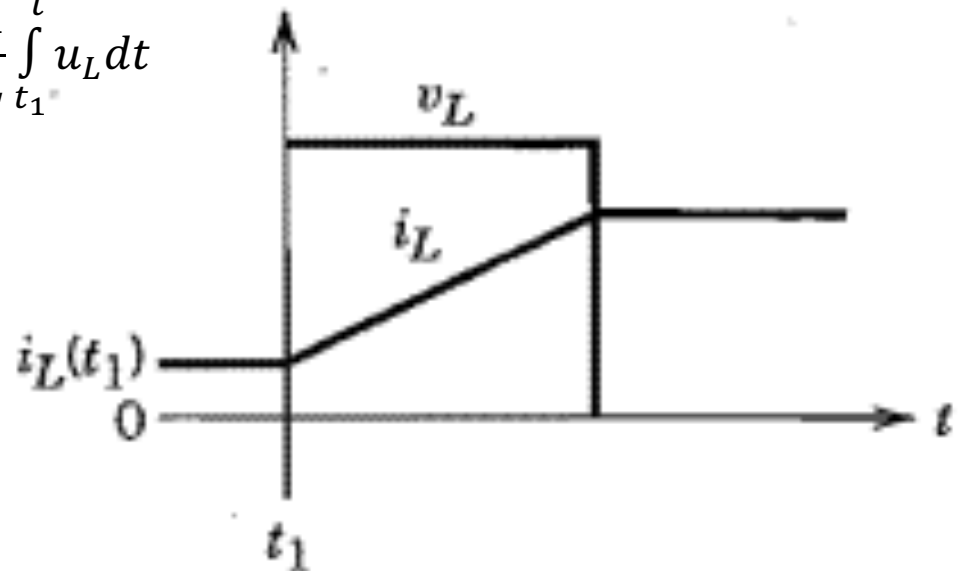
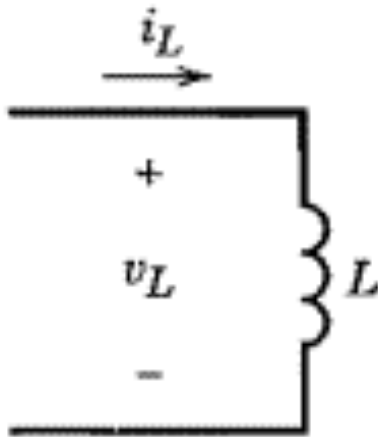
Inductor model

- Time and frequency domain

- $u_L = L \frac{di_L}{dt}$ $\bar{U}_L = j\omega L \bar{I}_L$



$$i_L(t) = i_L(t_1) + \frac{1}{L} \int_{t_1}^t u_L dt$$



Continuous conduction mode, cont.

- i_L never zero

- Steady state $\Rightarrow A = B$

$$t_{on}(V_d - V_o) = V_o(T_s - t_{on}) \Rightarrow \frac{V_o}{V_d} = \frac{t_{on}}{T_s} = D$$

- Average v_L zero in steady state

- Average v_{oi} output voltage,

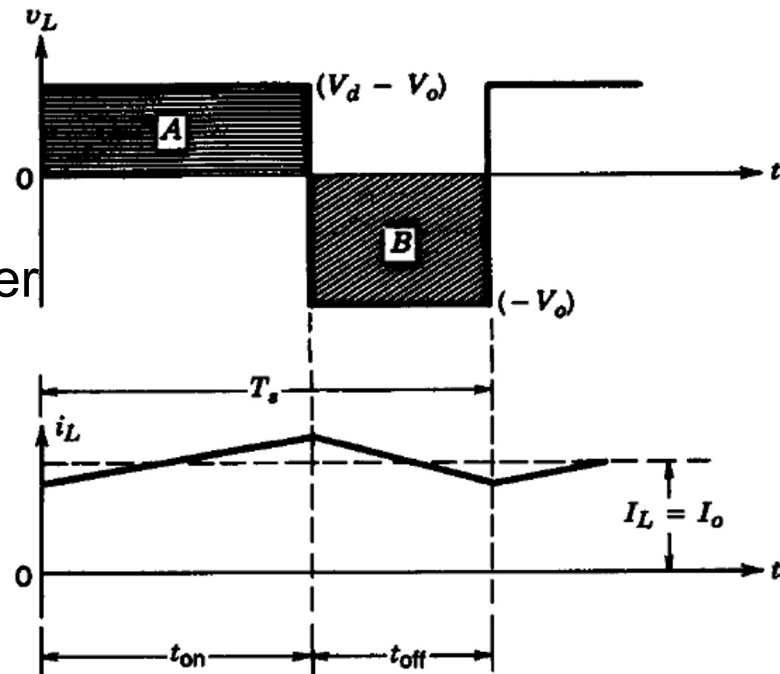
$$V_o = \frac{V_d t_{on} + 0 \cdot t_{off}}{T_s} = V_d D$$

- Ideal conditions: No power loss in converter

$$P_d = P_o \Rightarrow V_d I_d = V_o I_o$$

$$\frac{I_o}{I_d} = \frac{V_d}{V_o} = \frac{1}{D}$$

- DC transformer with turns ratio equal to D



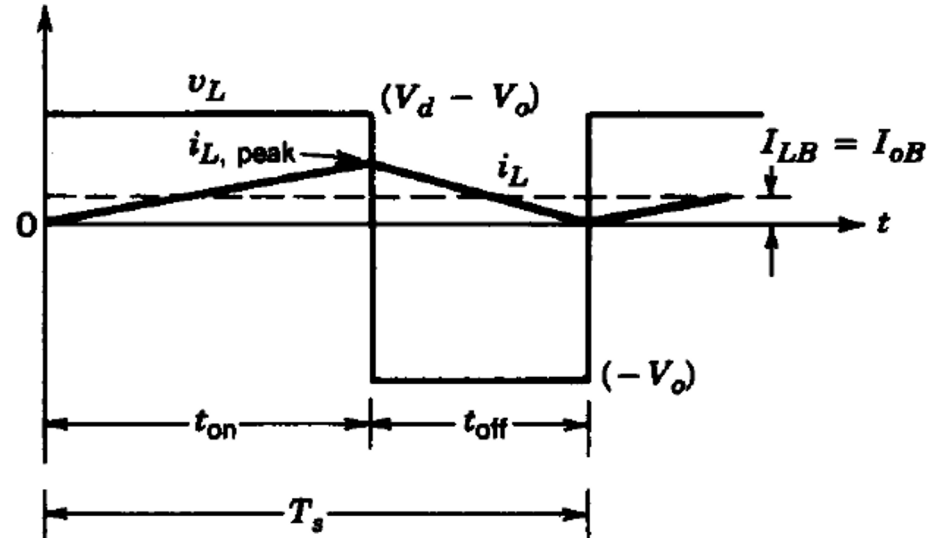
Discontinuous/Continuous Conduction mode boundary

- i_L reach zero at end of period
- Average inductor current equals the output current I_o
- Boundary current, I_{LB} , is the minimum average inductor current in **Cont.** mode

$$I_{LB} = \frac{1}{2} i_{L,peak} = \frac{t_{on}}{2L} (V_d - V_o)$$

$$I_{LB} = \frac{DT_s}{2L} (V_d - V_o)$$

$$I_{LB} = \frac{V_o}{2Lf_s} (1 - D)$$



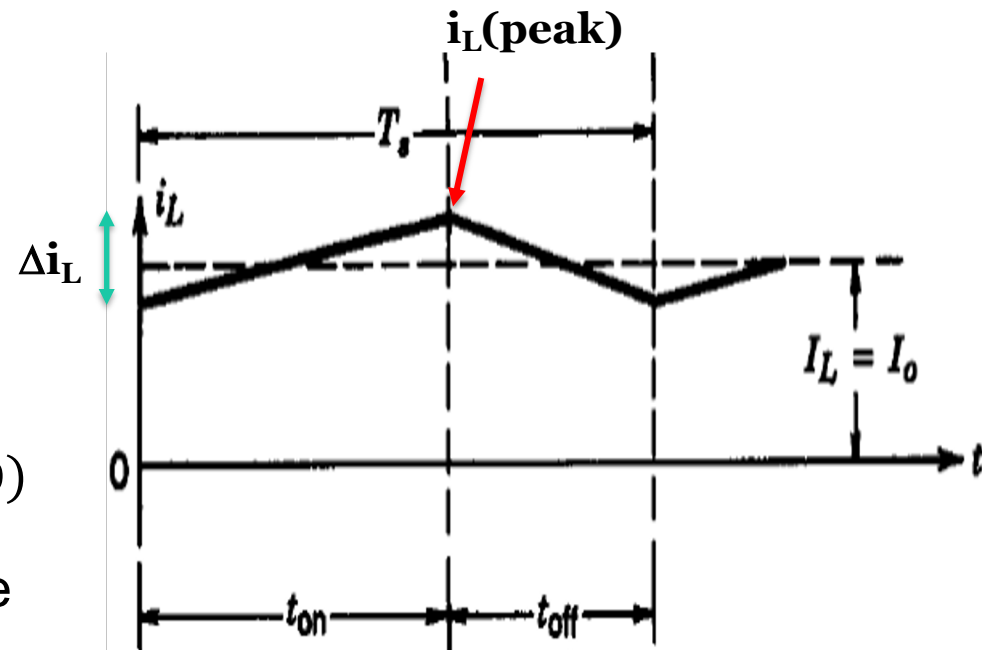
Inductance selection

- Inductance defines $\frac{di_L}{dt}$
- Max inductor current

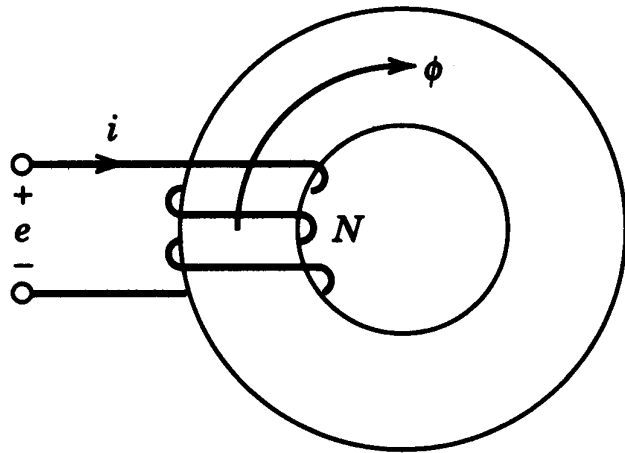
$$i_{L,peak} = I_o + \frac{\Delta i_L}{2}$$

$$\Delta i_L = \frac{V_o}{L f_s} (1 - D)$$

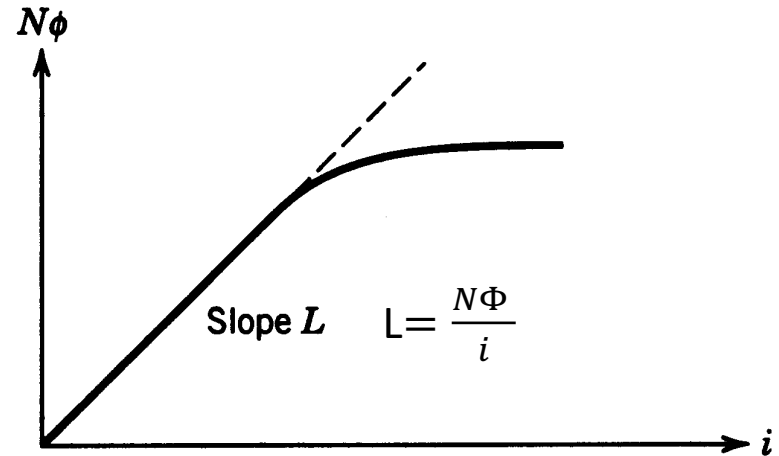
- Inductance $L = \frac{V_o}{\Delta i_L f_s} (1 - D)$
- Minimum D is the worse case



Inductance L



(a)



(b)

Figure 3-17 Self-inductance L .

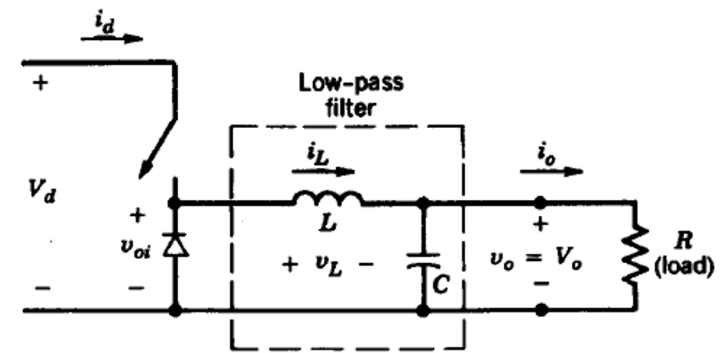
- Faraday's voltage induction law: $e = N \frac{d\Phi}{dt} = L \frac{di}{dt}$
- Magnetic saturation, $B = \frac{\Phi}{A} > B_{max}$

7-1

In a step-down converter, consider all components to be ideal. Let $v_o \approx V_o$ be held constant at 5 V by controlling the switch duty ratio D .

- Calculate the minimum inductance L required to keep the converter operation in a continuous-conduction mode under all conditions if:

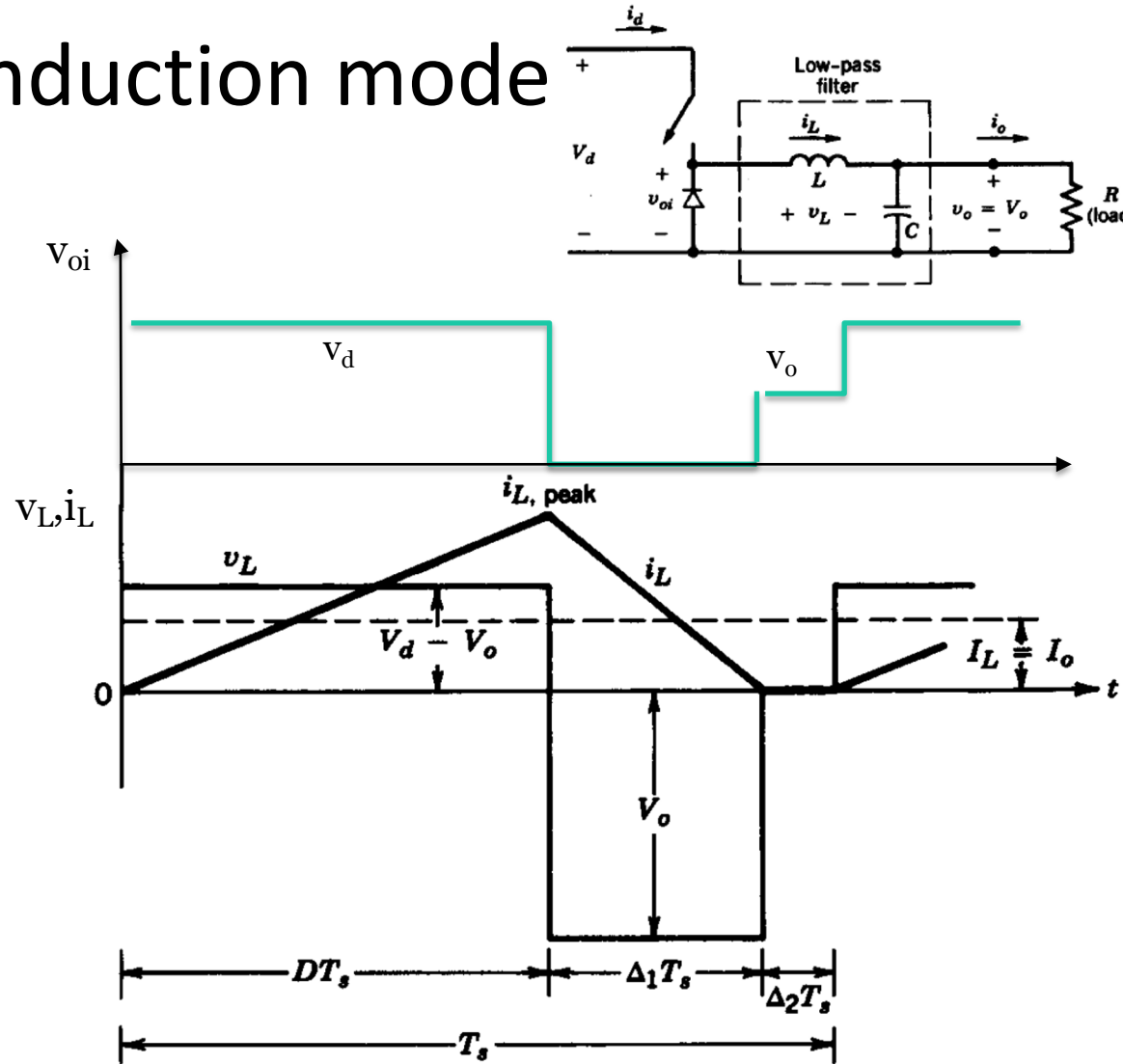
V_d is 10-40 V, $P_o \geq 5$ W, and $f_s = 50$ kHz.



Discontinuous conduction mode

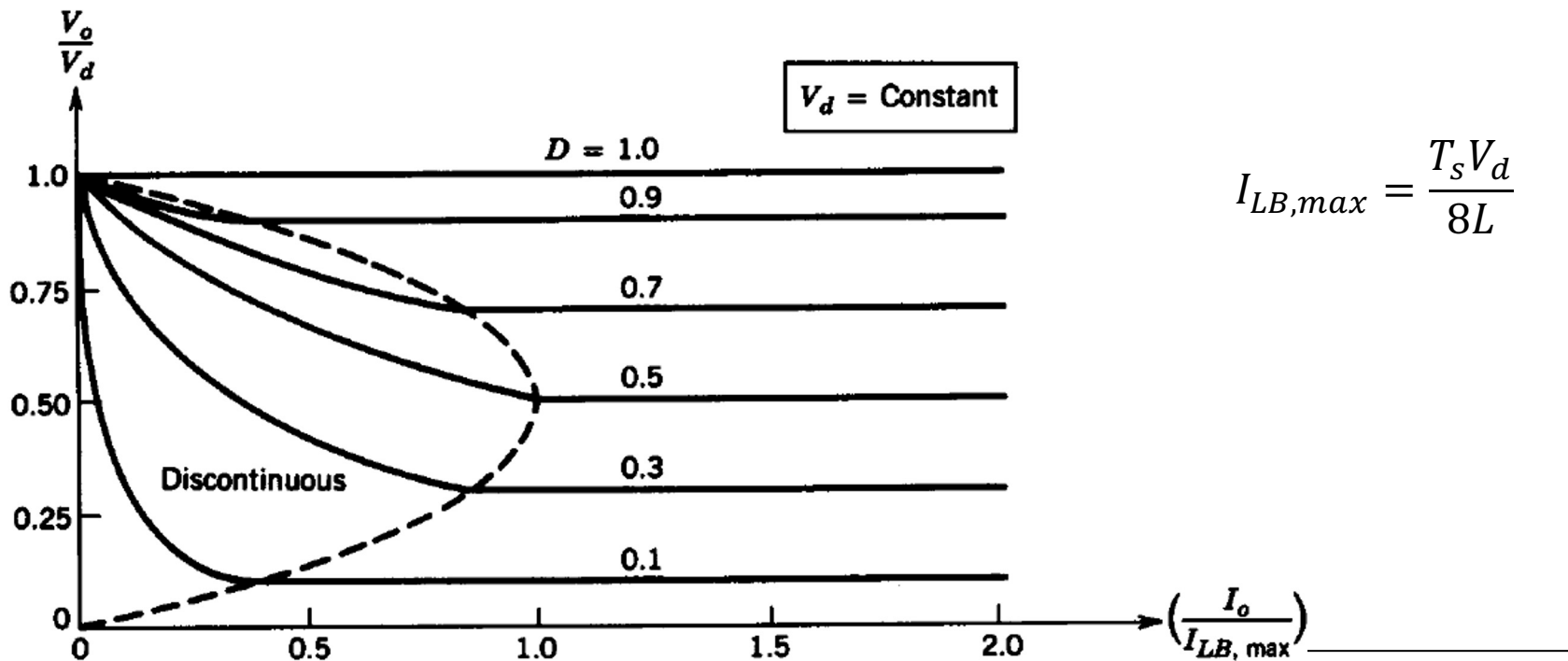
- $V_{oi} = V_d$ when switch on
- When switch off
 - $V_{oi} = 0$ if $i_L > 0$
 - $V_{oi} = V_o$ if $i_L = 0$
- Output vs. input voltage relation more complex

$$\frac{V_o}{V_d} = \frac{D^2}{D^2 + \frac{1}{4} (I_o / I_{LB,max})}$$



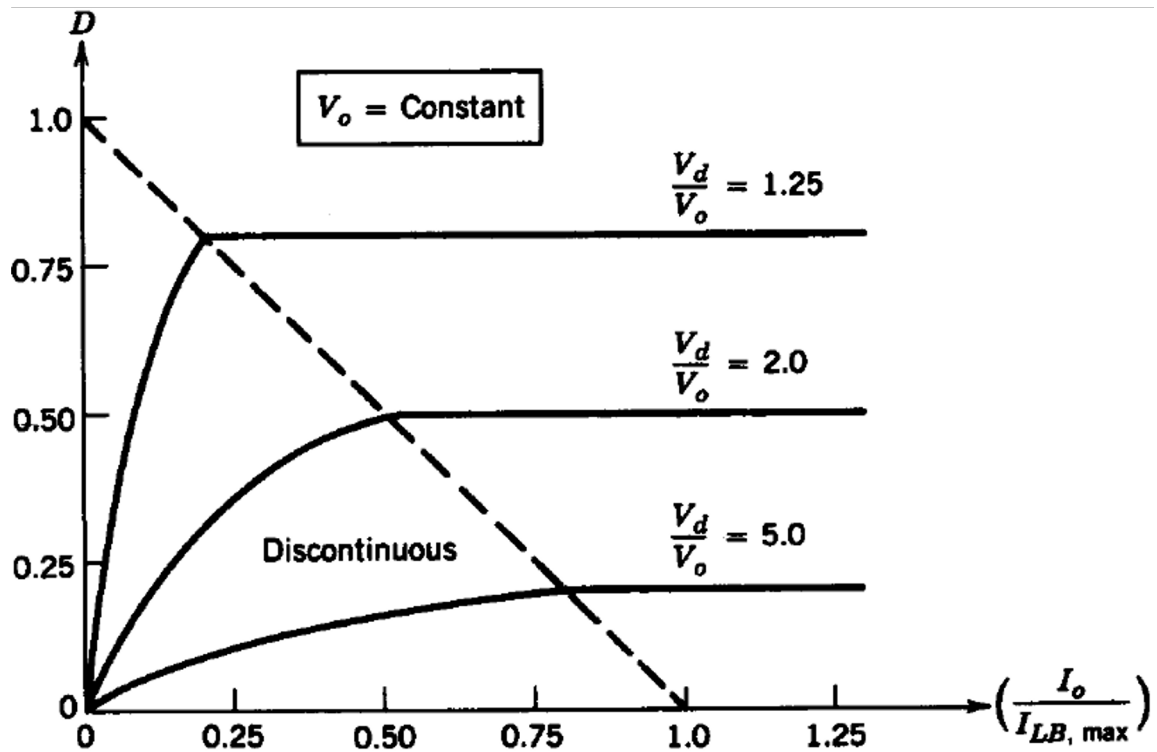
Constant V_d step-down characteristic

- Very low load result in increased output voltage!



Discontinuous conduction mode with constant V_o

- $I_{LB,max} = \frac{T_s V_o}{2L}$
- Control ratio for constant V_o



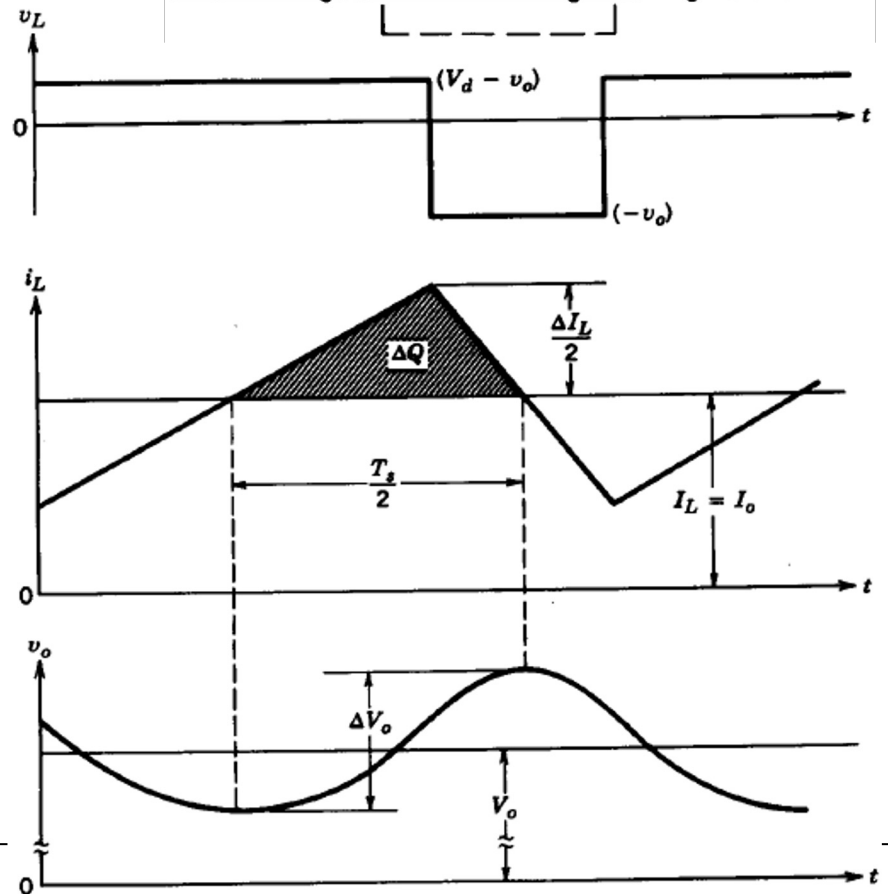
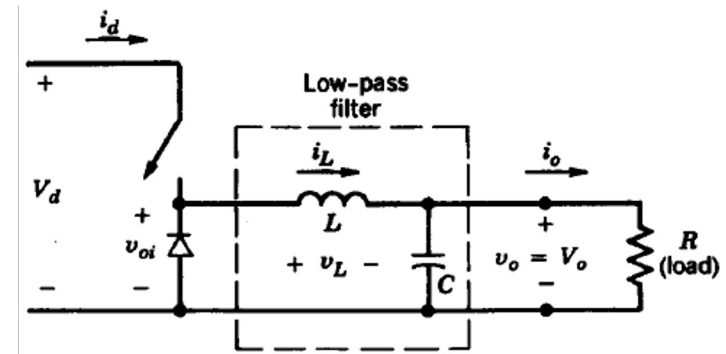
Output voltage ripple

- Assuming:
 - Continuous current
 - Inductor ripple current in C
 - Average inductor current in R

$$\Delta V_o = \frac{\Delta Q}{C} = \frac{1}{C} \frac{1}{2} \frac{\Delta I_L T_s}{2}$$

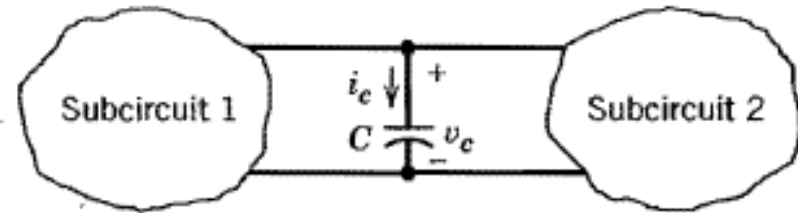
$$\Delta I_L = \frac{V_o}{L} (1 - D) T_s$$

$$\frac{\Delta V_o}{V_o} = \frac{1}{8} \frac{T_s^2 (1 - D)}{LC}$$

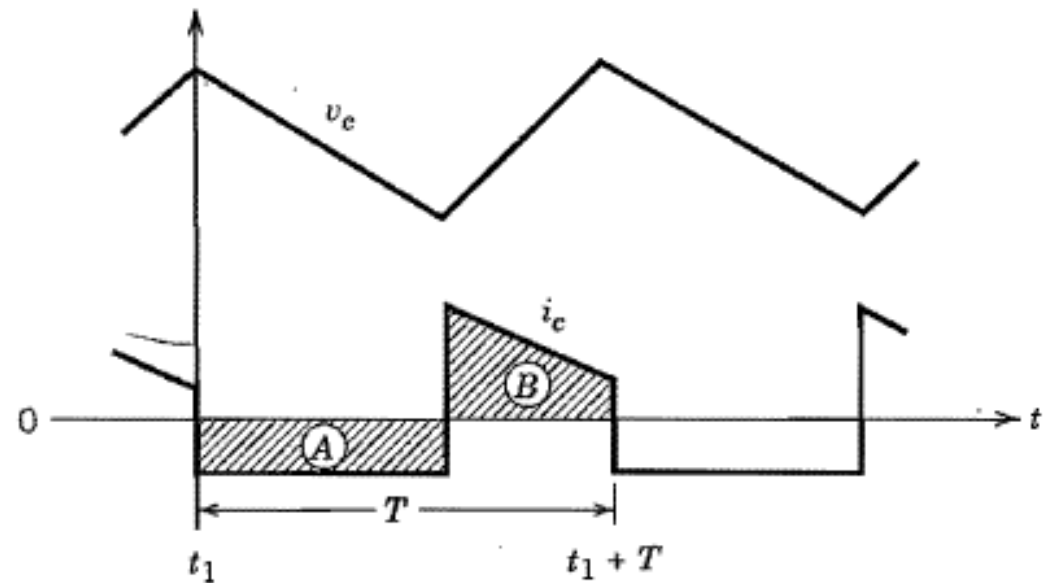


Capacitor a voltage storage

- Current is required to change the voltage
- $i_c > 0$: u_c increase
- $i_c < 0$: u_c decrease
- Steady state:
 - Constant average voltage
 - Current-time area (charge) $A=B$
 - Average current=0

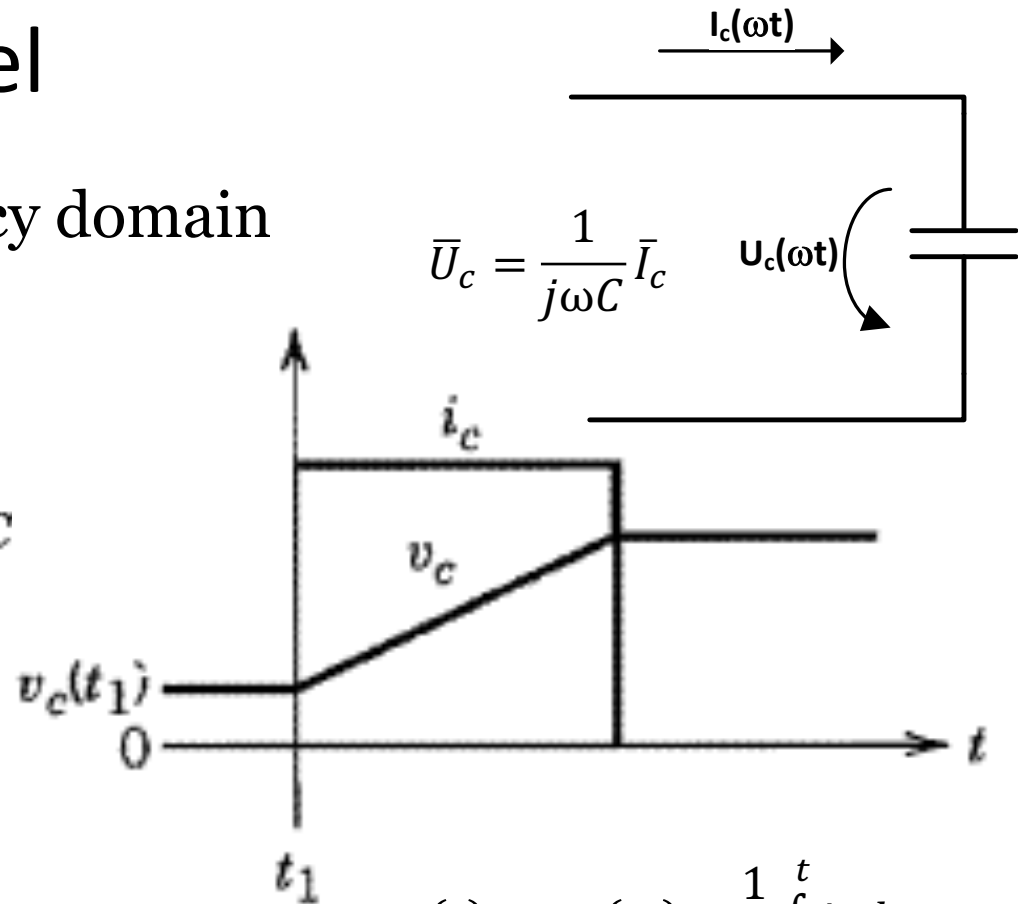
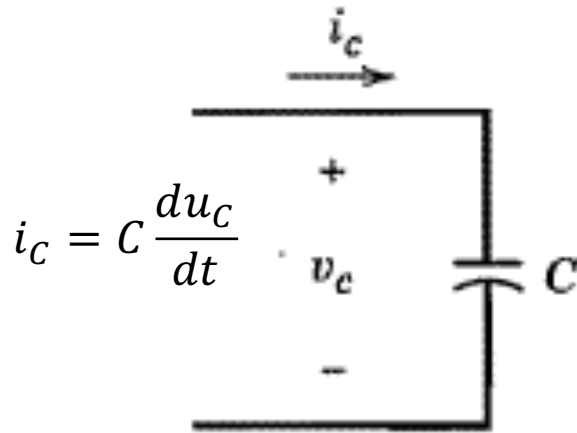


(a)



Capacitor model

- Time and frequency domain



Step-down (Buck) converter summary

- Output vs input

$$\frac{V_o}{V_d} = D$$

$$\frac{I_o}{I_d} = \frac{1}{D}$$

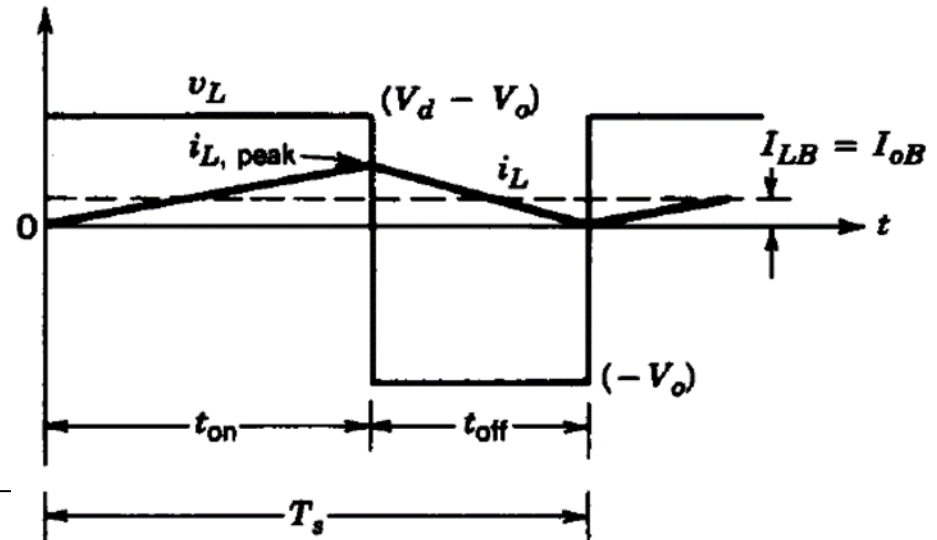
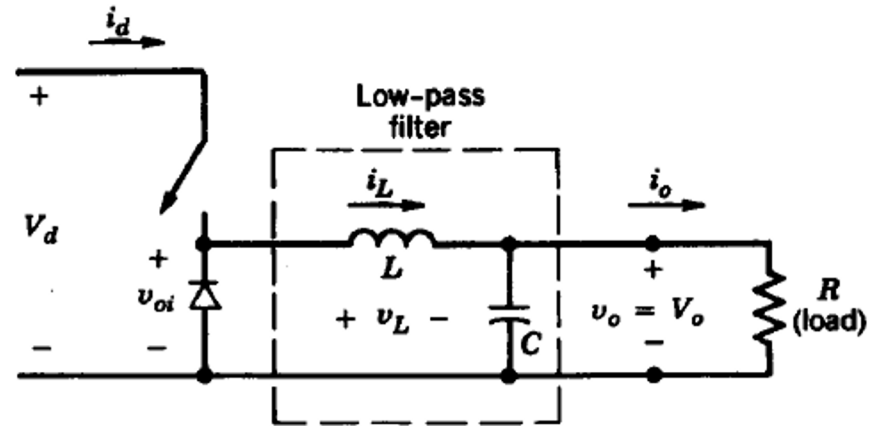
- High ripple current

$$\Delta i_L = \frac{V_o}{Lf_s} (1 - D)$$

- Output ripple voltage

$$\frac{\Delta V_o}{V_o} = \frac{1}{8} \frac{T_s^2 (1 - D)}{LC}$$

- Practical for D not lower than 0.2

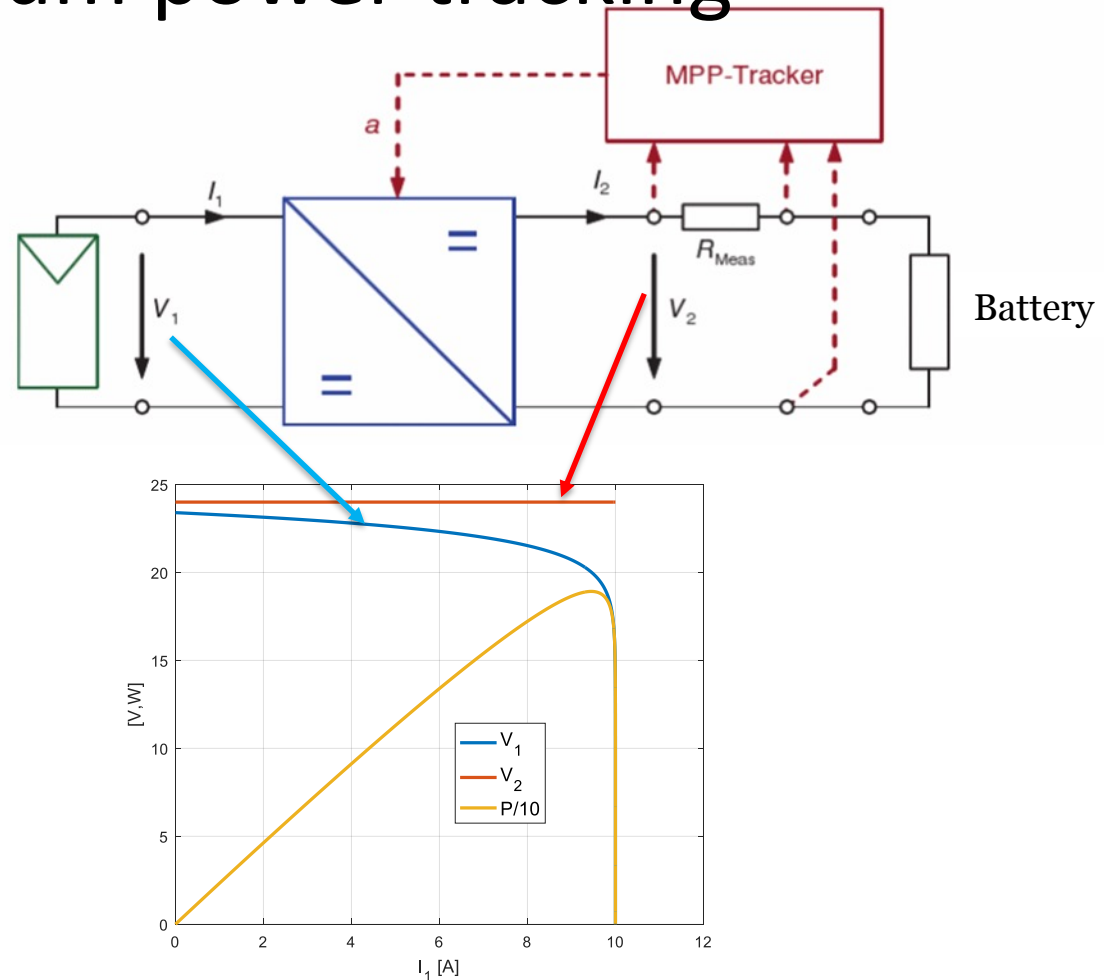


Lecture 2

DC/DC (Boost) up converter

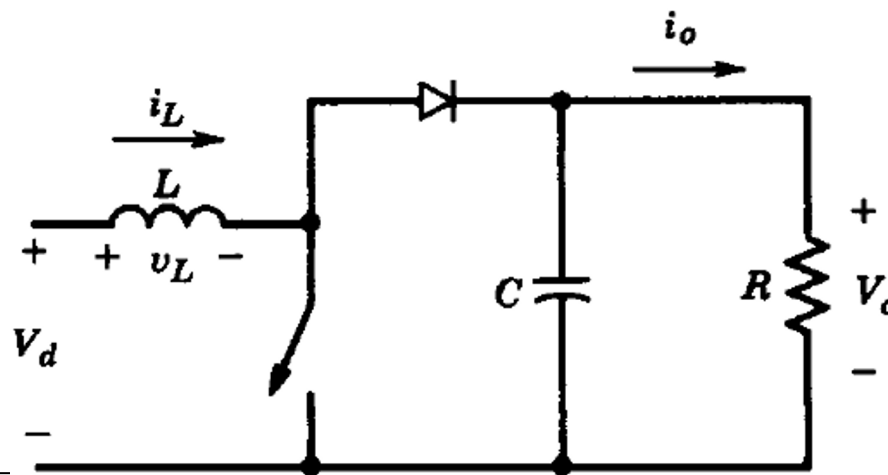
PV-panel maximum power tracking

- DC/DC conversion to optimize PV-panel voltage for maximum power
- Constant (24V) across battery load
- Variable PV-panel voltage (0-23 V)



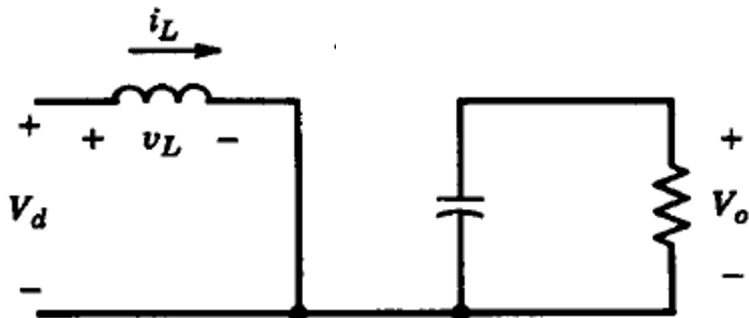
Step-up (Boost) converter

- Output must be larger than input voltage
 - Otherwise V_d is driving V_o directly through the diode $\Rightarrow V_o = V_d$
- Load energy into inductor, then output energy into load while still consuming energy from source
- C large enough to give low ripple, $v_o(t) \approx V_o$

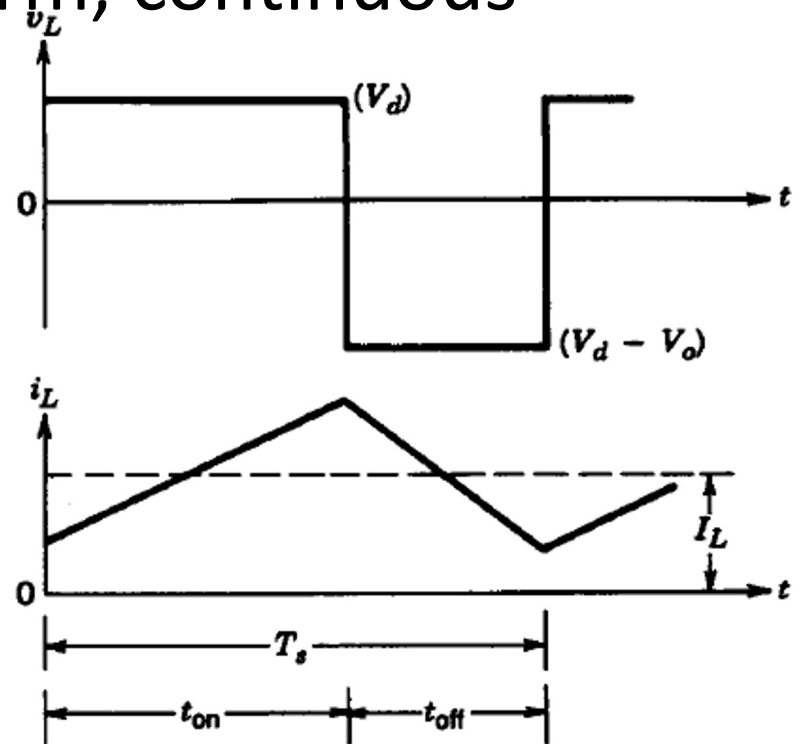
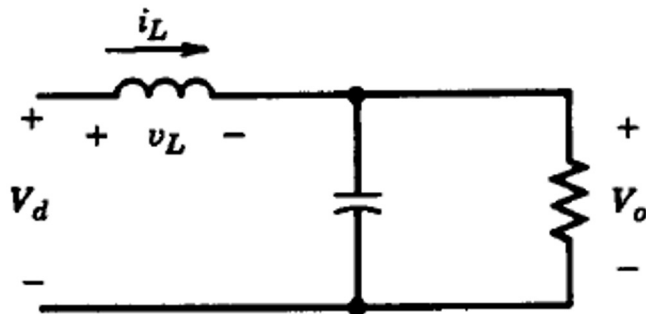


Step-up converter waveform, continuous conduction mode

- Switch on, diode off



- Switch off, diode on



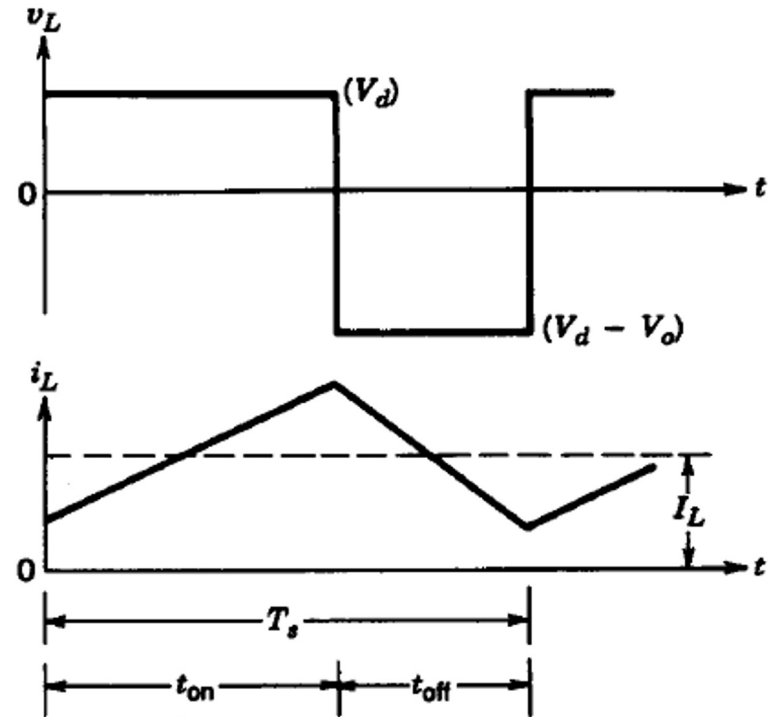
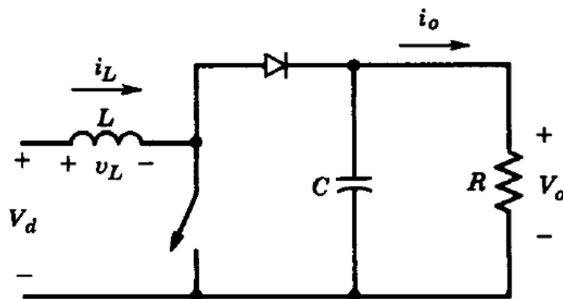
Step-up converter, continuous mode

$$V_d t_{on} + (V_d - V_o) t_{off} = 0$$

$$V_o/V_d = \frac{T_s}{t_{off}} = \frac{1}{1-D}$$

Lossless circuit: $V_d I_L = V_o I_o$

$$\Rightarrow \frac{I_o}{I_L} = (1-D)$$



Boundary between continuous and discontinuous mode

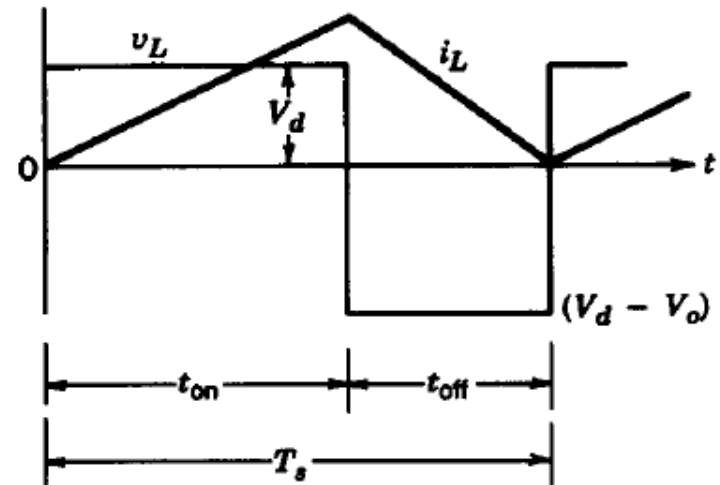
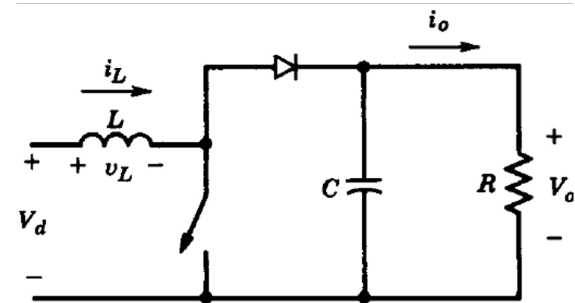
- Inductor current boundary

$$I_{LB} = \frac{1}{2} i_{L,peak} = \frac{1}{2} \frac{t_{on}}{L} V_d$$

$$I_{LB} = \frac{T_s D}{2L} V_o (1 - D)$$

- Output current boundary

$$I_{oB} = \frac{T_s D}{2L} V_o (1 - D)^2$$



Step-up converter, inductance choice

Continuous current boundary:

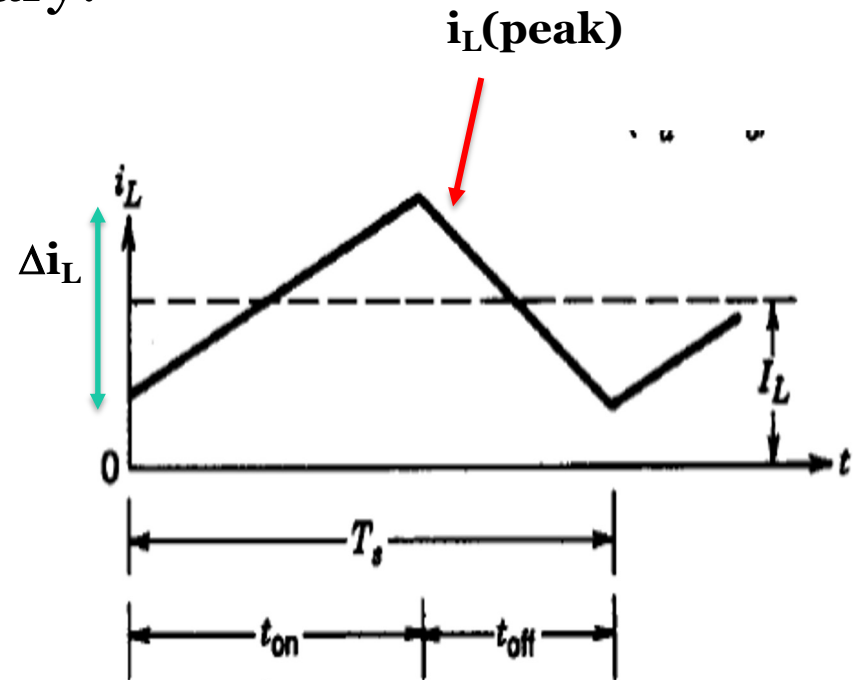
- Max inductor current

$$i_{L,peak} = I_L + \frac{\Delta i_L}{2}$$

$$\Delta i_L = \frac{DV_o}{Lf_s} (1 - D)$$

- Inductance

$$L = \frac{DV_o}{\Delta i_L f_s} (1 - D)$$

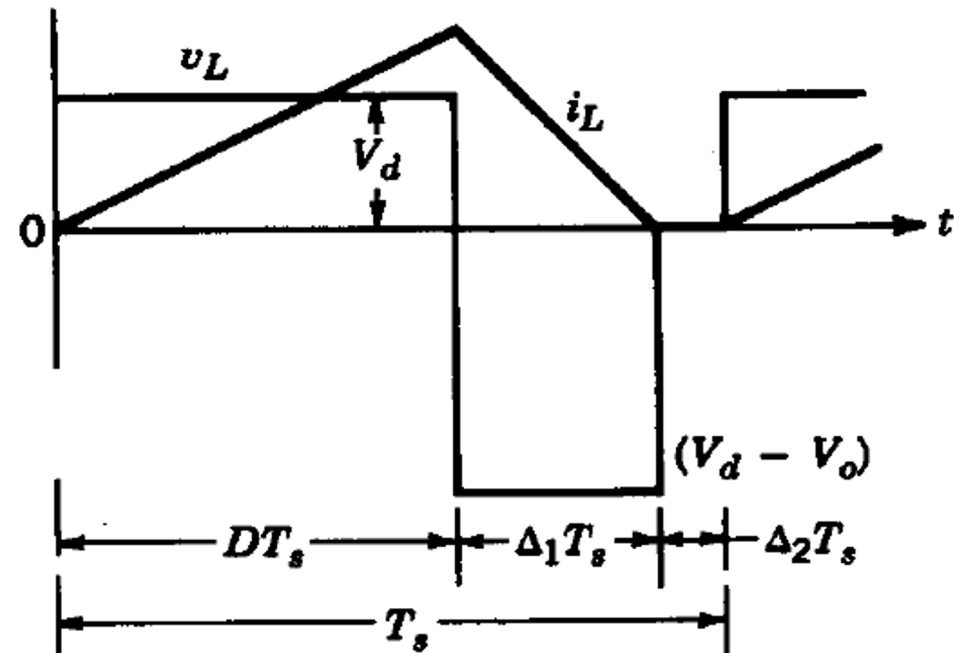
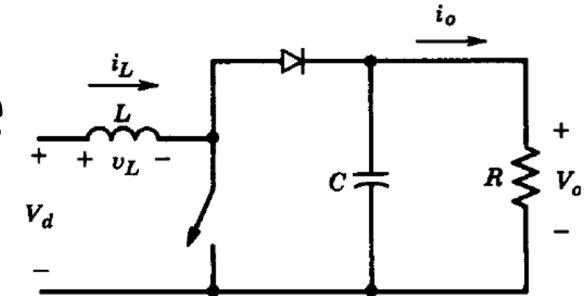


Step-up, discontinuous mode

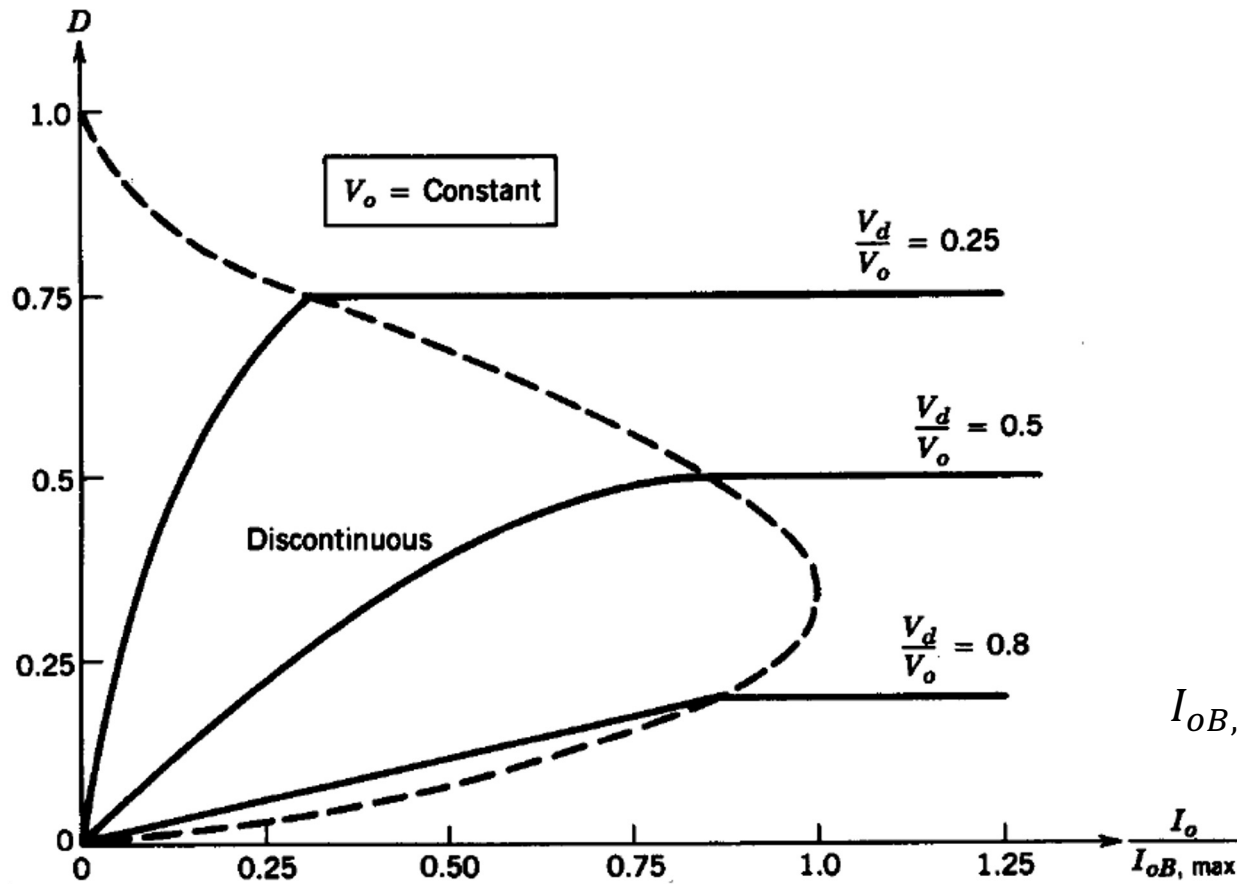
Comparing with continuous mode for same duty cycle

- Current I_L up ramp same
- Current I_L down ramp shorter due to current zero
- Higher di_L/dt
- Higher V_o for equal area in inductor voltage
- How to control D in discontinuous mode

$$D = \left[\frac{4}{27} \frac{V_o}{V_d} \left(\frac{V_o}{V_d} - 1 \right) \frac{I_o}{I_{oB,max}} \right]^{1/2}$$



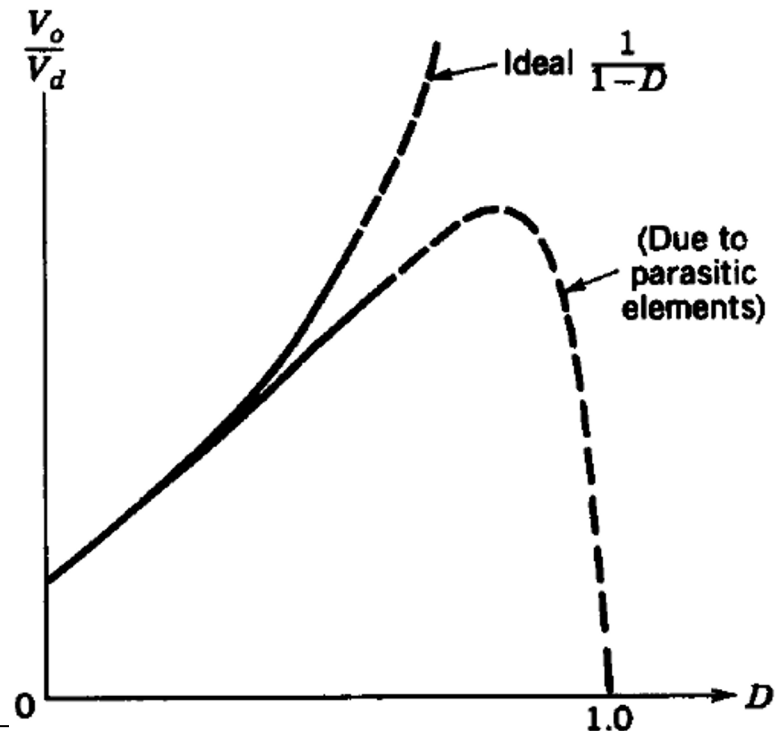
Step up converter characteristics, V_o constant



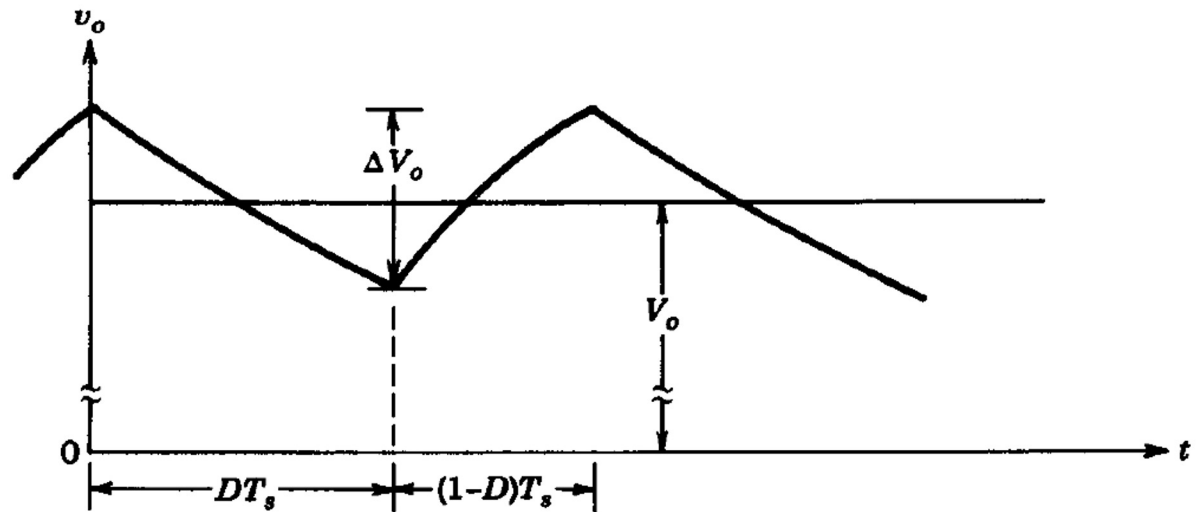
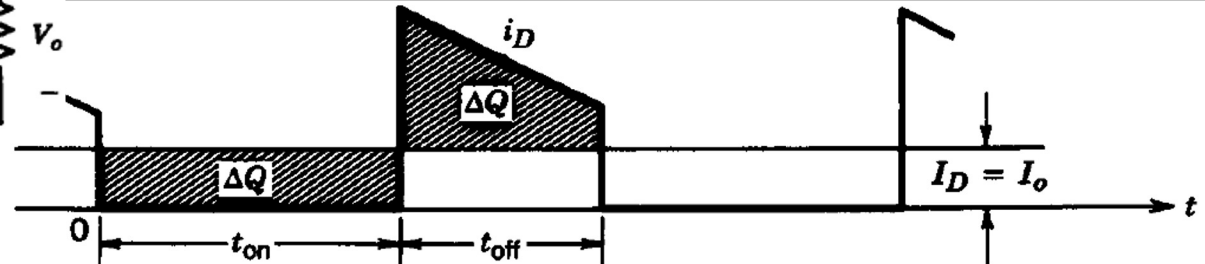
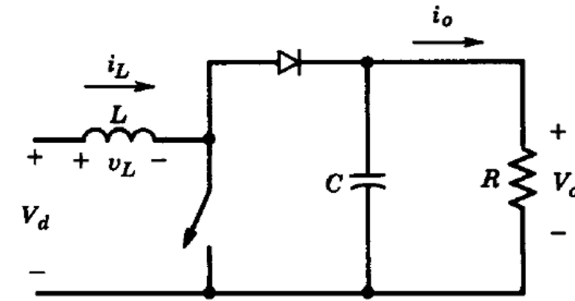
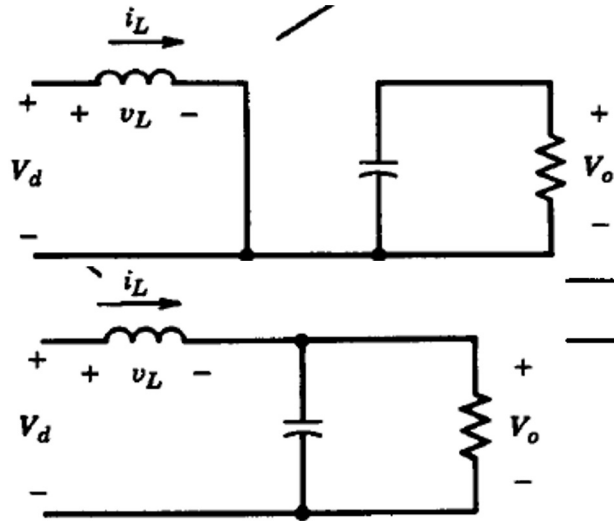
$$I_{oB, \max} = 0.074 \frac{T_s V_o}{L}$$

Effects of parasitics

- Losses in L, diode, switch, C
- Limited also by acceptable D ratio (approx. <3)



Output voltage ripple



$$\Delta V_o = \frac{\Delta Q}{C} = \frac{I_o D T_s}{C}$$

$$\frac{\Delta V_o}{V_o} = \frac{D T_s}{RC} = D \frac{T_s}{\tau}$$

where

$$\tau = RC \text{ time constant}$$

$$\min(i_L) > 0$$

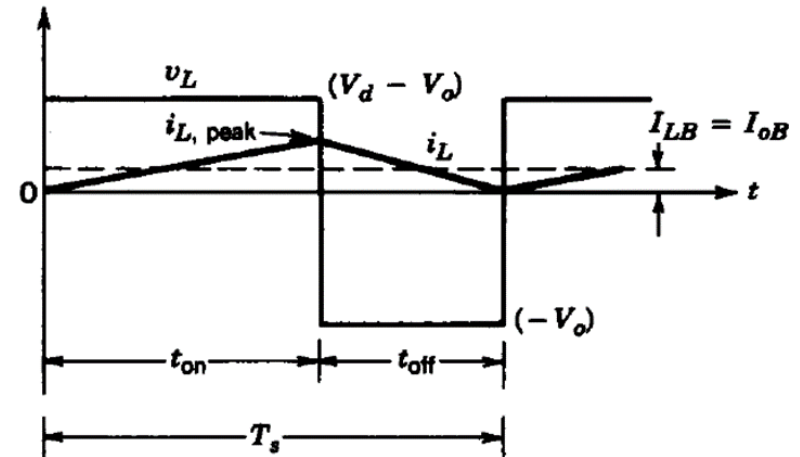
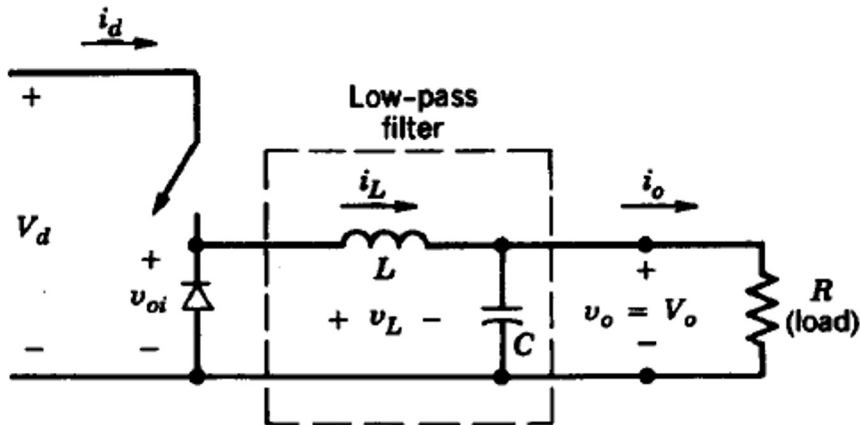
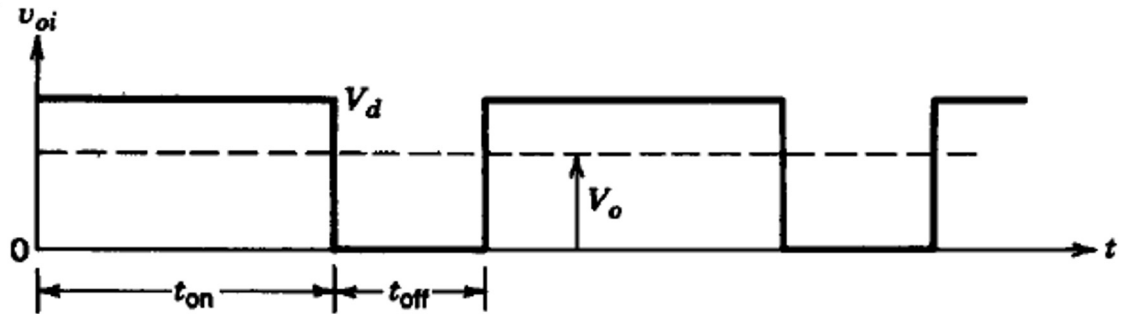
Lecture 2

Buck/Boost DC/DC Converter

Step-down (Buck) converter

- Output voltage lower than input voltage

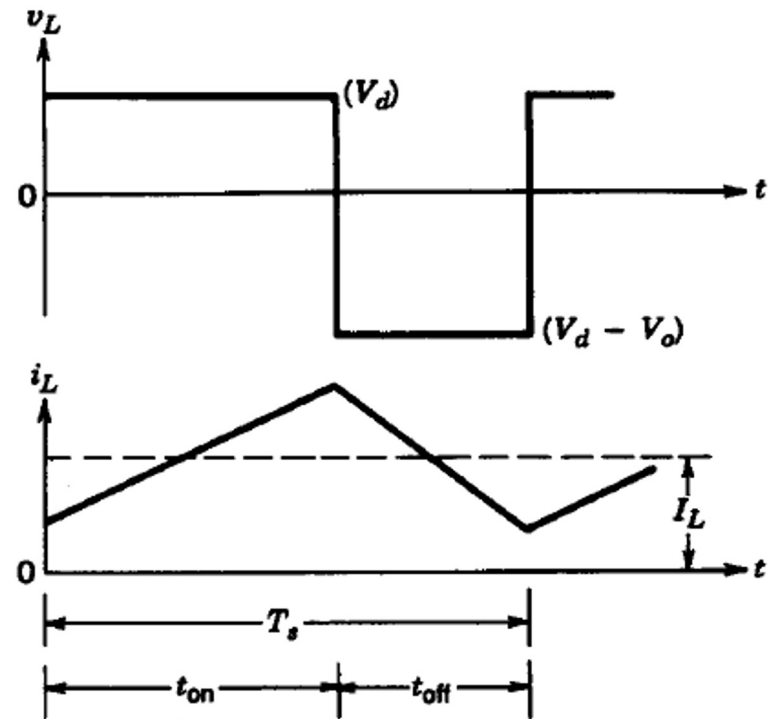
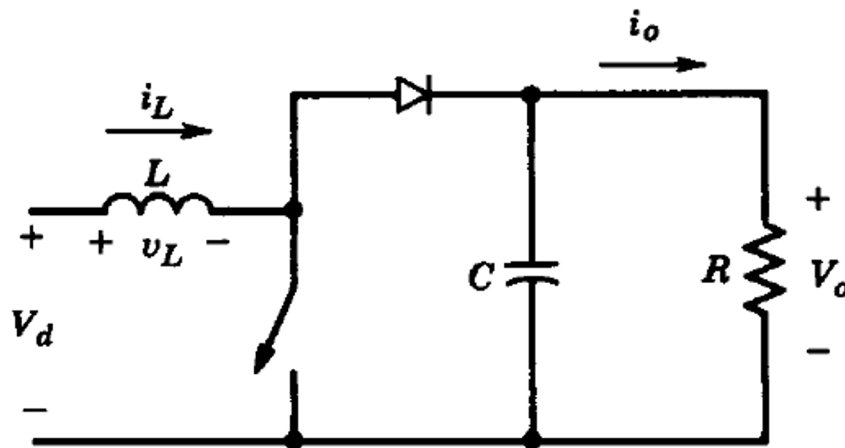
$$\frac{V_o}{V_d} = D$$



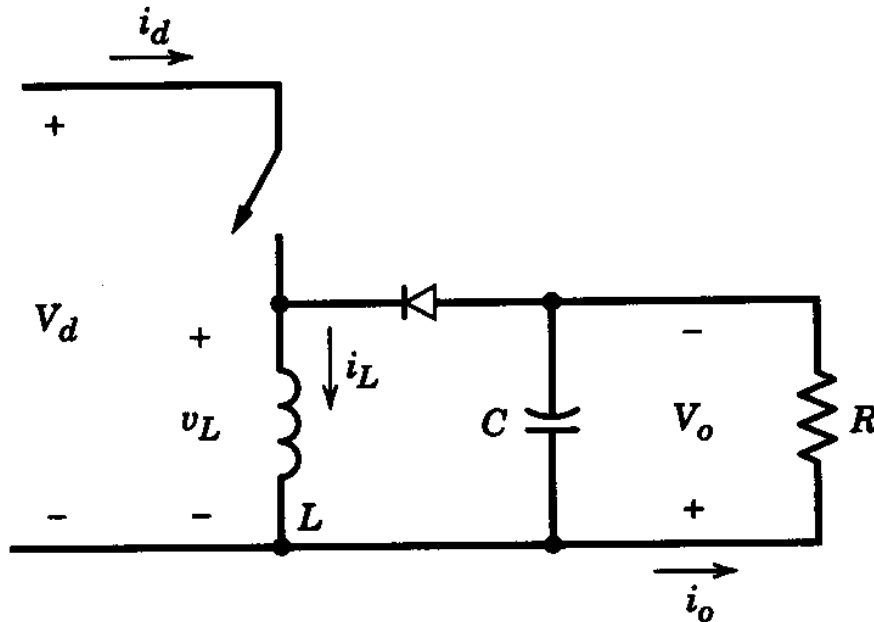
Step-up (boost) converter

- Output must be larger than input voltage

- $\frac{V_o}{V_d} = \frac{T_s}{t_{off}} = \frac{1}{1-D}$



Step-Down/Up (buck-boost) DC-DC Converter



$$\frac{V_o}{V_d} = D \frac{1}{1 - D}$$

Figure 7-18 Buck–boost converter.

- Negative output voltage
- The output voltage can be higher or lower than the input voltage

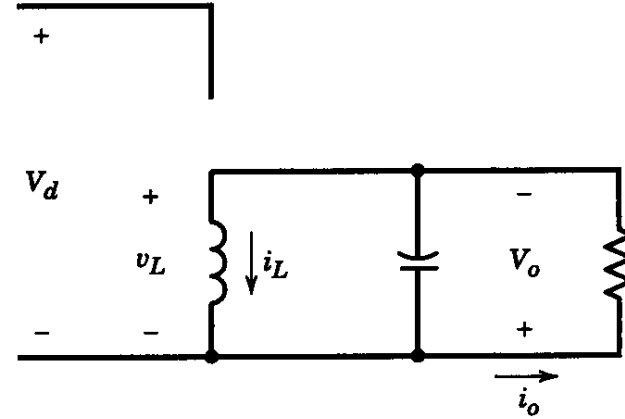
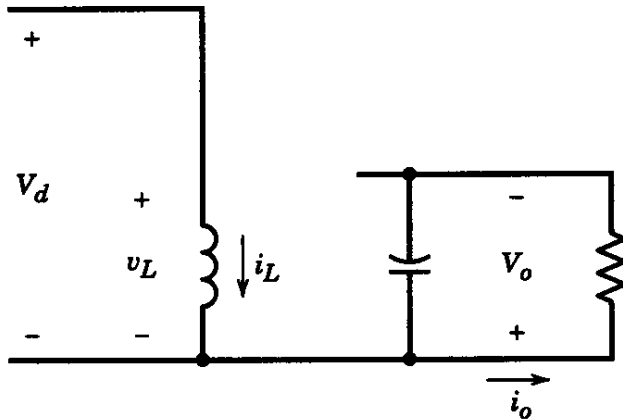
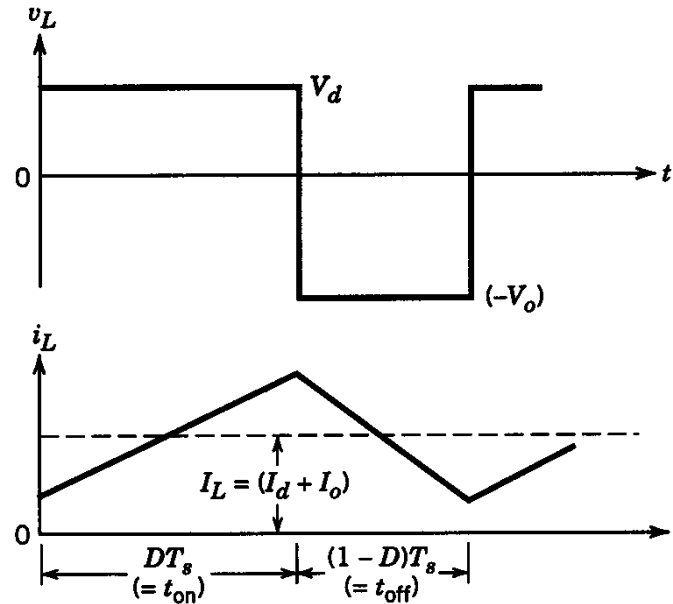
Buck-boost DC-DC Converter: Waveforms

Buck-boost converter

$$i_L > 0$$

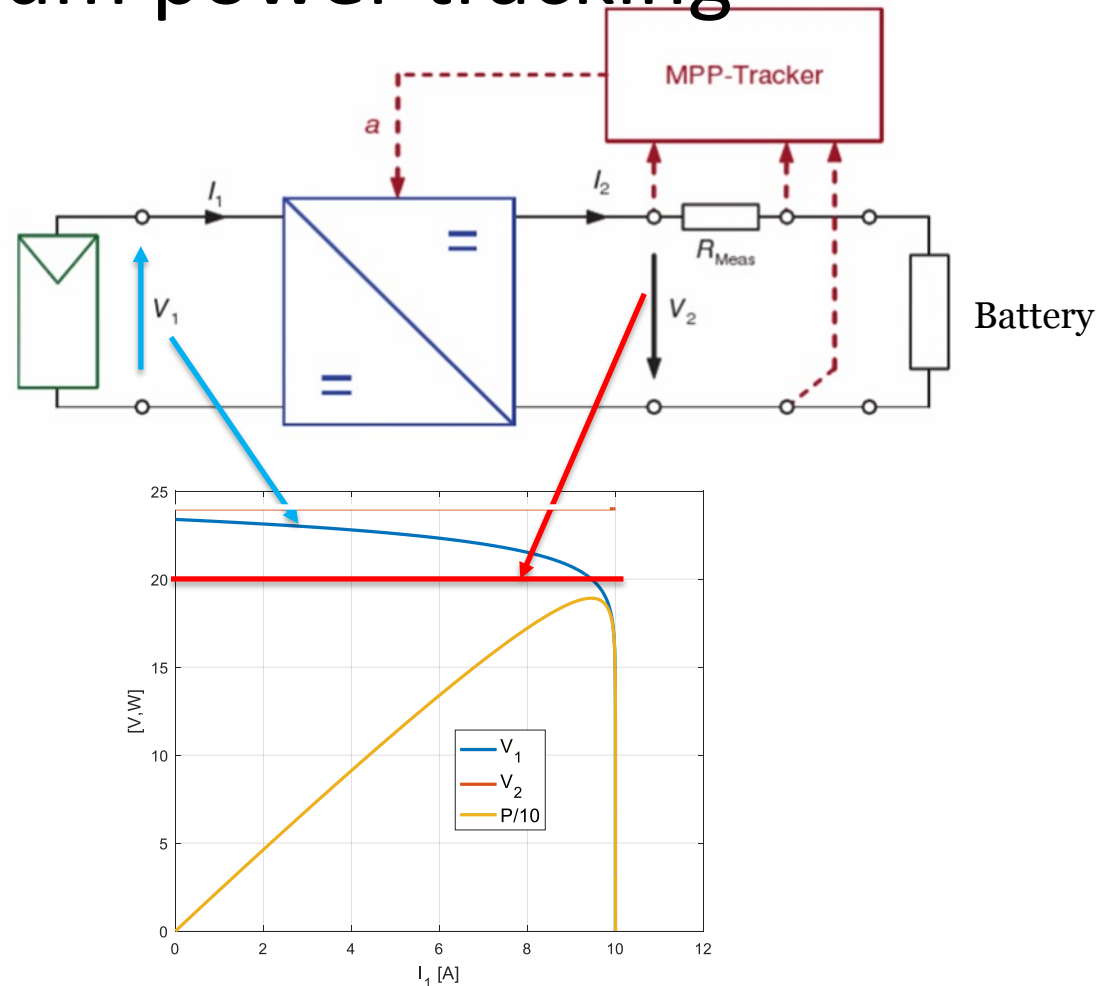
a) Switch on

b) Switch off



PV-panel maximum power tracking

- DC/DC conversion to optimize PV-panel voltage for maximum power
- Constant (20 V) across load
- Variable PV-panel voltage (0-23 V)
- Reversed PV-polarity



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